

Supervisor IC

System Power Good + Watchdog Timer + Reset for Automotive

BD39042MUF-C

General Description

BD39042MUF-C is a supervisor IC with quad power good, Watchdog timer and reset. This IC enables existing system to improve its ASIL level easily. BD39042MUF-C includes built-in self-test (BIST).

Features

- AEC-Q100 Qualified^(Note 1)
- Functional Safety Supportive Automotive Products
- Quad Power Good for External Inputs
- Over Voltage Detection (OVD)
- Under Voltage Detection (UVD)
- Adjustable Window Watchdog Timer(WDT)
- Reset for VDD Input (POR)
- Built-in Self-test (BIST)
- (Note 1) Grade 1

Applications

- Automotive for ADAS
- Camera Module
- Microwave Module
- Power Train ECU
- Other ECU

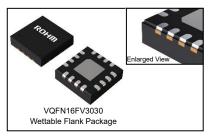
Key Specifications

- VDD Input Voltage Range: 2.7 V to 5.5 V (VDD voltage level needs to be fixed within this range in 6 % accuracy to avoid RSTIN reset detection)
- Detection Voltage (VDD POR / Power Good) Under Voltage Detection: -6 % (1.4 % accuracy) Over Voltage Detection: +6 % (1.4 % accuracy) Reset Assertion Delay Time: 10 ms Power Good Assertion Delay Time: 10 ms
- Operating Temperature Range: -40 °C to +125 °C

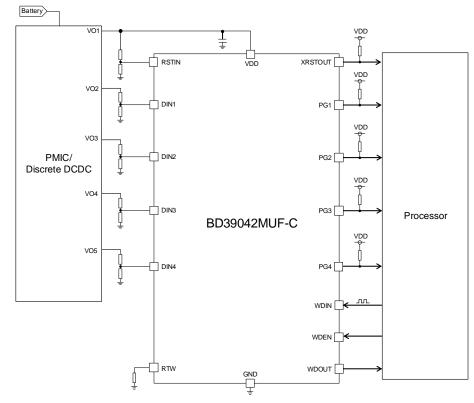
Package

VQFN16FV3030

W (Typ) x D (Typ) x H (Max) 3.0 mm x 3.0 mm x 1.0 mm



Typical Application Circuit

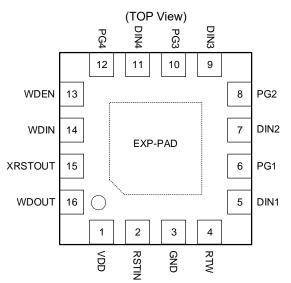


OProduct structure : Silicon integrated circuit OThis product has no designed protection against radioactive rays.

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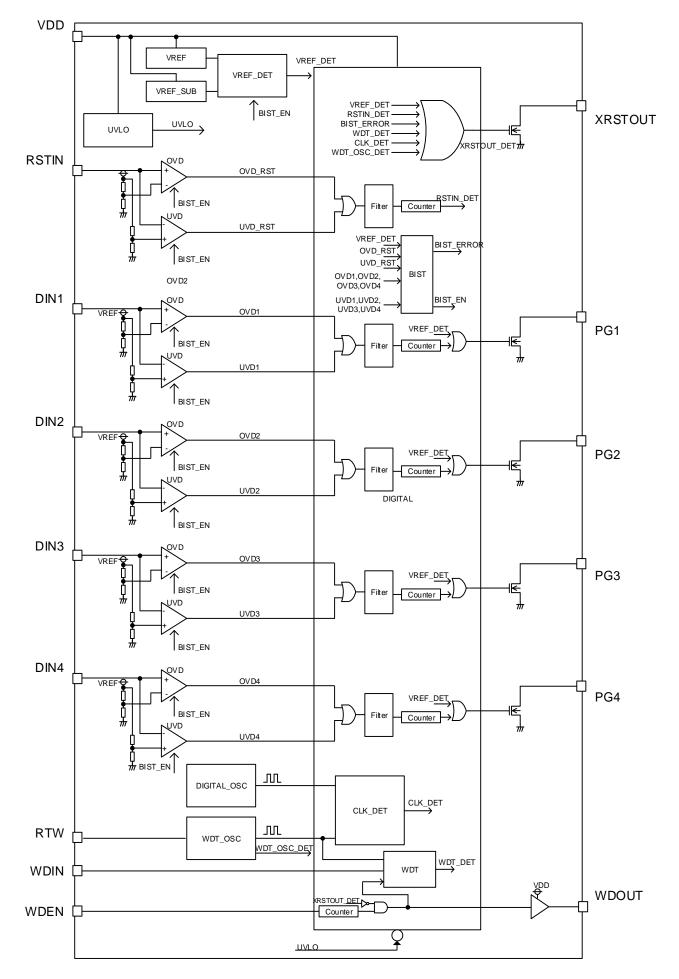
Pin Configuration



Pin Descriptions

Descriptions	D 1 D 1	
Pin No.	Pin Name	Function
1	VDD	IC's Power Source
2	RSTIN	The VDD pin voltage divided by external resistor input pin. Normal voltage level needs to be 0.8 V.
3	GND	IC's Power Ground
4	RTW	WDT frequency setting pin. FAST Timeout and SLOW Timeout is adjusted by the resistor value for this pin.
5	DIN1	Voltage for monitoring channel divided by external resistor input pin. Normal voltage level needs to be 0.8 V.
6	PG1	POWER GOOD output pin for the DIN1 pin, and Nch Open Drain output. Hi-Z for normal and Low for abnormal (reset). Please be pulled-up by external resistor. It can be pulled-up to any voltage source.
7	DIN2	Voltage for monitoring channel divided by external resistor input pin. Normal voltage level needs to be 0.8 V.
8	PG2	POWER GOOD output pin for the DIN2 pin, and Nch Open Drain output. Hi-Z for normal and Low for abnormal (reset). Please be pulled-up by external resistor. It can be pulled-up to any voltage source.
9	DIN3	Voltage for monitoring channel divided by external resistor input pin. Normal voltage level needs to be 0.8 V.
10	PG3	 POWER GOOD output pin for the DIN3 pin, and Nch Open Drain output. Hi-Z for normal and Low for abnormal (reset). Please be pulled-up by external resistor. It can be pulled-up to any voltage source.
11	DIN4	Voltage for monitoring channel divided by external resistor input pin. Normal voltage level needs to be 0.8 V.
12	PG4	POWER GOOD output pin for the DIN4 pin, and Nch Open Drain output. Hi-Z for normal and Low for abnormal (reset). Please be pulled-up by external resistor. It can be pulled-up to any voltage source.
13	WDEN	Enable pin for WDT. High = Active, Low = Disable and WDT error is ignored.
14	WDIN	Clock input pin for WDT
15	XRSTOUT	Reset output pin. Nch Open Drain output. Hi-Z for normal and Low for abnormal (reset). Please be pulled-up by external resistor. It can be pulled-up to any voltage source. Either error of OVD, UVD for RSTIN, reference voltage monitoring, internal OSC monitoring, WDT and BIST at power- up sequence causes this pin to drive low.
16	WDOUT	Buffer output pin for the WDEN pin input. Abnormal Power Source / the GND pin shortage for the WDEN pin can be recognized by monitoring this pin. This pin becomes Low when the XRSTOUT pin is low.
-	EXP-PAD	The EXP-PAD is connected to the PCB Ground plane.

Block Diagram



Block Diagrams - continued

Description of Blocks Reference Voltage (VREF) VREF is reference voltage of each input voltage.

Reference Voltage (VREF_SUB) VREF_SUB is reference voltage of VREF_DET.

Reference Voltage (VREF_DET)

This monitors the 2 reference voltage, VREF and VREF_SUB. This block contributes to high reliability by continuously and mutually monitoring correct operation. Occurrence of error leads to Low output at the XRSTOUT pin and which is never de-asserted as long as abnormal status lasts. It becomes High at 10 ms (Typ) after the voltage returned to the normal range.

Under Voltage Lockout Circuit (UVLO)

Protection circuit to prevent internal circuit from malfunction at lower voltage (Power-up sequence or input power supply drop). This monitors the VDD pin voltage and UVLO works when it goes down to threshold level.

As UVLO is detected, the XRSTOUT, WDOUT, PG1, PG2, PG3 and PG4 pins output Low. Also Counter value in DIGITAL BLOCK is initialized and DIGITAL_OSC/WDT_OSC stop working.

Oscillator (DIGITAL_OSC) This OSC generates the clock to control DIGITAL BLOCK. The frequency of DIGITAL_OSC is fixed at 2.2 MHz.

Oscillator (WDT_OSC)

This OSC generates the clock to control WDT.

The frequency of WDT_OSC is possible to be adjusted by the resistor value, so that FAST Timeout / SLOW Timeout is changed by that.

WDT_OSC has the function to stop its working when the external resistor at the RTW pin is shorted or OPEN (WDT_OSC_DET). Once CLK_DET is detected, XRSTOUT becomes Low.

Oscillator (CLK_DET)

This block monitors both DIGITAL_OSC and WDT_OSC. 2 OSCs always monitor their frequency each other and it leads to the higher reliability.

When an error happened at the monitoring, XRSTOUT becomes Low.

Over Voltage Detection (OVD1, OVD2, OVD3, OVD4, OVD RST)

When input voltage goes over the threshold level, OVD is detected and the PG1, PG2, PG3 and PG4 pins are driven by Low. Detecting pins are DIN1, DIN2, DIN3 and DIN4 and RSTIN. OVD detection for the DIN1, DIN2, DIN3 and DIN4 pins causes corresponding the PG1, PG2, PG3 and PG4 pins to become Low. OVD detection for RSTIN causes the XRSTOUT pin to become Low. These output signals become High at 10 ms (Typ) after each input pin returns within the normal voltage range. And each input has a filter in DIGITAL BLOCK, then overshoot within 50 µs (Min) is ignored.

Under Voltage Detection (UVD1, UVD2, UVD3, UVD4, UVD_RST)

When input voltage goes below the threshold level, UVD is detected and the PG1, PG2, PG3 and PG4 pins are driven by Low. Detecting pins are DIN1, DIN2, DIN3, DIN4 and RSTIN. UVD detection for the DIN1, DIN2, DIN3 and DIN4 pins causes corresponding the PG1, PG2, PG3 and PG4 pins to become Low. UVD detection for RSTIN causes the XRSTOUT pin to become Low. These output signals become High at 10 ms (Typ) after each input pin returns within the normal voltage range. And each input has a filter in DIGITAL BLOCK, then undershoot within 50 µs (Min) is ignored.

BIST

When VDD Power on Reset (monitoring the RSTIN pin) is released, BIST is performed and self-test for DIN1, DIN2, DIN3, DIN4, RSTIN and VREF_DET comparators are executed to see if each comparator correctly toggles their High/Low output based on input level change.

BIST time (t_{BIST}) is 2 ms (Max). Once BIST ends without any errors, XRSTOUT becomes High. If an error is found during BIST, XRSTOUT keeps Low and BIST is repeated until it passes.

Watchdog Timer (WDT)

Watchdog Timer (WDT) monitors microprocessor's operation by detecting the time from both rise and fall edge of WDIN. If BIST result is abnormal, WDT does not work and XRSTOUT is kept low. WDT is activated when WDOUT = High, and both WDEN and XRSTOUT have to be High in order to get WDOUT to be High.

As long as the duty of WDIN clock is kept within "Trigger open window" in Figure 1, WDT does not detect any errors and XRSTOUT stays at High. Sequence for FAST Timeout and SLOW Timeout are shown in Figure 2 and Figure 3.

Description of Blocks - continued

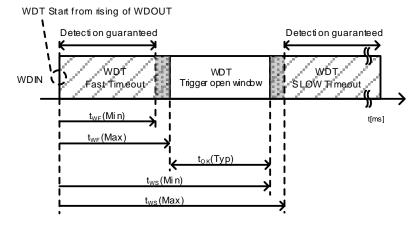


Figure 1. WDT Window Description

WDT FAST Timeout Detection

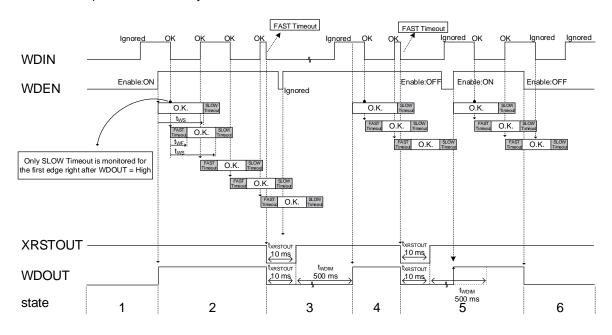
- 1. WDIN input signal is ignored when WDOUT = Low. WDT is activated when WDOUT = High, and both WDEN and XRSTOUT have to be High in order to get WDOUT to be High.
- 2. For the initial duration just after WDOUT goes to High, only SLOW Timeout detection works and FAST Timeout does not work. Either Low or High input to the WDIN pin is acceptable as initial level. Once rising-up or falling-down edge of WDIN comes within SLOW Timeout, both FAST Timeout and SLOW Timeout detections start to work.
- 3. These time detection monitors the time until next edge and when it detects WDIN edge within FAST Timeout (twF), XRSTOUT and WDOUT becomes Low. XRSTOUT goes back to High after 10 ms (Typ) delay, while WDOUT goes back to High after twDIM (500 ms: Typ) in addition to txRSTOUT (10 ms: Typ) as long as WDEN = High. twDIM is implemented as a time for microprocessor to be reset normally and stabilized. If this time is unnecessary and WDT should be activated as soon as possible, WDEN may be controlled like state 5 in the Figure 2.

WDEN toggle during XRSTOUT = Low is ignored.

4. When WDOUT becomes High, WDT is activated again and operation resumes.

Only SLOW Timeout detection works until the next first edge, and both SLOW Timeout and FAST Timeout starts at the first edge like state 1 in Figure 2.

- 5. If this time is unnecessary and WDT should be activated as soon as possible, WDEN may be controlled like in Figure 2. twDIM is canceled by toggling WDEN like High->Low->High and WDT is activated immediately even during twDIM. After WDT is enabled it works as same as state 2 in Figure 2.
- 6. When WDEN is Low, WDOUT becomes Low and WDT is disabled. During this period WDIN input signal is ignored and XRSTOUT output is not affected by that.





Block Diagrams - continued

WDT SLOW Timeout Detection

- 1. WDIN input signal is ignored when WDOUT = Low. WDT is activated when WDOUT = High, and both WDEN and XRSTOUT have to be High in order to get WDOUT to be High.
- For the initial duration just after WDOUT goes to High, only SLOW Timeout detection works and FAST Timeout does not work. Either Low or High input to the WDIN pin is acceptable as initial level. Once rising-up or falling-down edge of WDIN comes within SLOW Timeout, both FAST Timeout and SLOW Timeout detections start to work.
- These time detection monitors the time until next edge and when it cannot detect WDIN edge within SLOW Timeout (tws), XRSTOUT and WDOUT becomes Low. XRSTOUT goes back to High after 10 ms (Typ) delay, while WDOUT goes back to High after twDIM (500 ms: Typ) in addition to txRSTOUT (10 ms: Typ) as long as WDEN = High. twDIM is implemented as a time for microprocessor to be reset normally and stabilized. If this time is unnecessary and WDT should be activated as soon as possible, WDEN may be controlled like state 5 in the Figure 3.

WDEN toggle during XRSTOUT = Low is ignored.

- 4. When WDOUT becomes High, WDT is activated again and operation resumes.
- Only SLOW Timeout detection works until the next first edge, and both SLOW Timeout and FAST Timeout starts at the first edge like state 1 in Figure 3.
- 5. If this time is unnecessary and WDT should be activated as soon as possible, WDEN may be controlled like the Figure 3. twDIM is canceled by toggling WDEN like High->Low->High and WDT is activated immediately even during twDIM. After WDT is enabled it works as same as state 2 in Figure 3.
- 6. When WDEN is Low, WDOUT becomes Low and WDT is disabled. During this period WDIN input signal is ignored and XRSTOUT output is not affected by that.

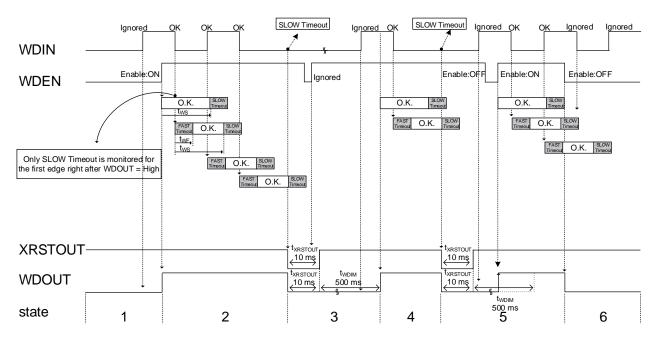


Figure 3. WDT SLOW Timeout Detection

WDT SLOW Timeout Detection - continued

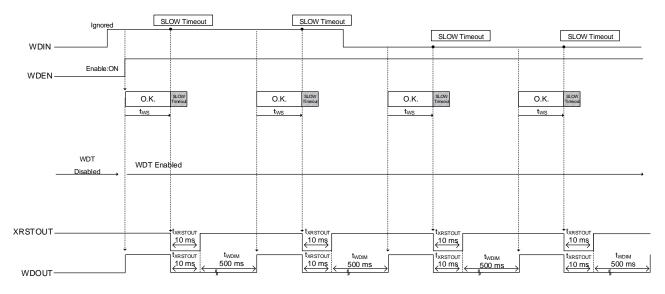


Figure 4. XRSTOUT Behavior with Continuous WDT Timeout Detected

The window time for detection can be changed by the resistor value between the RTW and GND pins. Following figure shows the detection time determined by R_{RTW} resistor value. Please refer to a table of electric characteristic regarding accuracy. Customer can choose the value ranging from 10 k Ω to 47 k Ω according to their clock frequency. The ratio for detection time is fixed and can be shown like this, FAST Timeout: SLOW Timeout = 1 : 2.

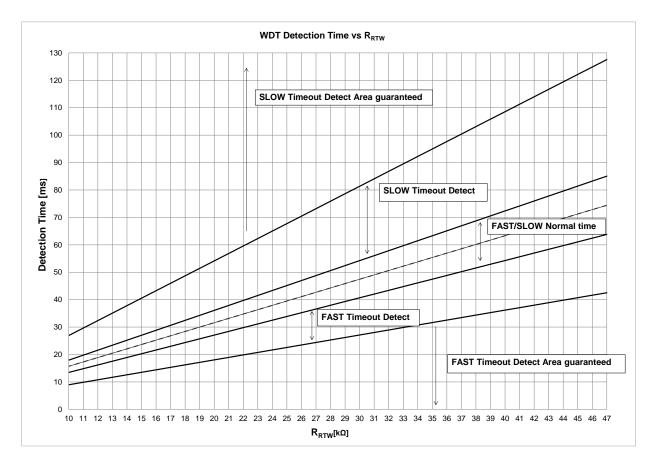


Figure 5. Detection Time vs R_{RTW}

Absolute Maximum Ratings

Parameter	Symbol	Rating	Unit
VDD Voltage	VDD	-0.3 to +7	V
RSTIN Voltage	V _{RSTIN}	-0.3 to +7	V
DIN1, DIN2, DIN3, DIN4 Voltage	Vdin1, Vdin2, Vdin3, Vdin4	-0.3 to +7	V
XRSTOUT Voltage	VXRSTOUT	-0.3 to +7	V
PG1, PG2, PG3, PG4 Voltage	G4 Voltage VPG1, VPG2, VPG3, VPG4		V
WDIN Voltage	Vwdin	-0.3 to +7	V
WDEN Voltage	Vwden	-0.3 to +7	V
WDOUT Voltage	Vwdout	-0.3 to V _{DD} +0.3 ^(Note 1)	V
RTW Voltage	V _{RTW}	-0.3 to V _{DD} +0.3 ^(Note 1)	V
Maximum Junction Temperature	Tjmax	Tjmax 150	
Storage Temperature Range	Tstg	-55 to +150	°C

Caution 1: Operating the IC over the absolute maximum ratings may damage the IC. The damage can either be a short circuit between pins or an open circuit between pins and the internal circuitry. Therefore, it is important to consider circuit protection measures, such as adding a fuse, in case the IC is operated over the absolute maximum ratings.

Caution 2: Should by any chance the maximum junction temperature rating be exceeded the rise in temperature of the chip may result in deterioration of the properties of the chip. In case of exceeding this absolute maximum rating, design a PCB with thermal resistance taken into consideration by increasing board size and copper area so as not to exceed the maximum junction temperature rating.

(Note 1) Maximum voltage is 7 V.

Thermal Resistance (Note 1)

Devenueter	Cumhal	Thermal Res	1.1	
Parameter	Symbol	1s ^(Note 3)	2s2p ^(Note 4)	Unit
VQFN16FV3030				
Junction to Ambient	θ _{JA}	189.0	57.5	°C/W
Junction to Top Characterization Parameter ^(Note 2)	Ψ_{JT}	23	10	°C/W

(Note 1) Based on JESD51-2A (Still-Air).

(Note 2) The thermal characterization parameter to report the difference between junction temperature and the temperature at the top center of the outside surface of the component package.

(Note 3) Using a PCB board based on JESD51-3. (Note 4) Using a PCB board based on JESD51-5, 7.

Layer Number of Measurement Board	Material	Board Size					
Single	FR-4	114.3 mm x 76.2 mm x	(1.57 mmt				
Тор							
Copper Pattern	Thickness						
Footprints and Traces	70 µm						
Layer Number of Measurement Board	Material	Board Size		Thermal V Pitch		e 5) Diameter	
4 Layers	FR-4	114.3 mm x 76.2 mm	114.3 mm x 76.2 mm x 1.6 mmt		Φ	Ф0.30 mm	
Тор		2 Internal Layers		Bottom			
Copper Pattern	Thickness	Copper Pattern	Thickness	Copper Pattern		Thickness	
Footprints and Traces	70 µm	74.2 mm x 74.2 mm	nm x 74.2 mm 35 µm 74.2 ı		m	70 µm	

(Note 5) This thermal via connects with the copper pattern of all layers.

Recommended Operating Conditions

Parameter	Symbol	Min	Тур	Max	Unit
Operating Ambient Temperature	Topr	-40	-	+125	°C
VDD Voltage	V _{DD}	2.7	-	5.5	V
WDIN Input Pulse Width	t _{WDIN}	10	-	125	ms
WDIN Minimum ON Pulse / OFF Pulse	twdp	-	-	100	μs

Electrical Characteristics (Unless otherwise specified V_{DD} = 2.7 V to 5.5 V, -40 °C \leq Ta \leq +125 °C)

Parameter	Symbol	Min	Тур	Max	Unit	Conditions
All		I	I		I	
Circuit Current	Ivdd	465	810	1345	μA	V _{DD} = 4.1 V, R _{RTW} = 27 kΩ, XRSTOUT, PG1, PG2, PG3, PG4 = H
VDD Power On Reset Threshold Voltage (Falling)	V _{VDDUV1}	2.25	2.50	2.65	V	VDD monitor
VDD Power On Reset Threshold Voltage (Rising)	V _{VDDUV2}	2.30	2.55	2.70	V	VDD monitor
VDD Power On Reset Hysteresis	VVDDHYS	-	50	-	mV	
Power Good						
DIN1 Power Good Low Detect Voltage	V _{UVD1}	0.7415	0.7520	0.7625	V	DIN1 Pin Voltage=Sweep down
DIN1 Power Good High Detect Voltage	V _{OVD1}	0.8362	0.8480	0.8598	V	DIN1 Pin Voltage=Sweep up
DIN1 Input Filter Time	t _{DIN1}	50	75	100	μs	
PG1 Low Voltage	V _{PG1L}	-	-	0.3	V	I _{PG1} = 1 mA
PG1 Leak Current	ILPG1	-	-	2	μA	V _{PG1} = 5.5 V
PG1 Assertion Delay Time	t _{PG1}	7	10	13	ms	
DIN2 Power Good Low Detect Voltage	VUVD2	0.7415	0.7520	0.7625	V	DIN2 Pin Voltage = Sweep down
DIN2 Power Good High Detect Voltage	V _{OVD2}	0.8362	0.8480	0.8598	V	DIN2 Pin Voltage = Sweep up
DIN2 Input Filter Time	t _{DIN2}	50	75	100	μs	
PG2 Low Voltage	VPG2L	-	-	0.3	V	I _{PG2} = 1 mA
PG2 Leak Current	ILPG2	-	-	2	μA	V _{PG2} = 5.5 V
PG2 Assertion Delay Time	t _{PG2}	7	10	13	ms	
DIN3 Power Good Low Detect Voltage	V _{UVD3}	0.7415	0.7520	0.7625	V	DIN3 Pin Voltage = Sweep down
DIN3 Power Good High Detect Voltage	V _{OVD3}	0.8362	0.8480	0.8598	V	DIN3 Pin Voltage = Sweep up
DIN3 Input Filter Time	t _{DIN3}	50	75	100	μs	
PG3 Low Voltage	Vpg3l	-	-	0.3	V	I _{PG3} = 1 mA
PG3 Leak Current	ILPG3	-	-	2	μA	V _{PG3} = 5.5 V
PG3 Assertion Delay Time	t _{PG3}	7	10	13	ms	
DIN4 Power Good Low Detect Voltage	V _{UVD4}	0.7415	0.7520	0.7625	V	DIN4 Pin Voltage = Sweep down
DIN4 Power Good High Detect Voltage	V _{OVD4}	0.8362	0.8480	0.8598	V	DIN4 Pin Voltage = Sweep up
DIN4 Input Filter Time	t _{DIN4}	50	75	100	μs	
PG4 Low Voltage	VPG4L	-	-	0.3	V	I _{PG4} = 1 mA
PG4 Leak Current	I _{LPG4}	-	-	2	μA	V _{PG4} = 5.5 V
PG4 Assertion Delay Time	t _{PG4}	7	10	13	ms	

Electrical Characteristics (Unless otherwise specified V_{DD}=2.7 V to 5.5 V, -40 °C \leq Ta \leq +125 °C) - continued

Parameter	Symbol	Min	Тур	Max	Unit	Conditions
VDD Power On Reset		L	I	L		
RSTIN Power Good Low Detect Voltage	VUVDRST	0.7415	0.7520	0.7625	V	RSTIN Pin Voltage = Sweep down
RSTIN Power Good High Detect Voltage	VOVDRST	0.8362	0.8480	0.8598	V	RSTIN Pin Voltage = Sweep up
RSTIN Input Filter Time	t RSTIN	50	75	100	μs	
XRSTOUT Low Voltage	VXRSTL	-	-	0.3	V	I _{XRSTOUT} = 1 mA
XRSTOUT Leak Current	IXRST	-	-	10	μA	V _{XRSTOUT} = 5.5 V
XRSTOUT Assertion Delay Time	txrstout	7	10	13	ms	
Watch Dog Timer						
FAST Timeout Detect1	t _{WF1}	9.0	11.2	13.5	ms	R _{RTW} = 10 kΩ
SLOW Timeout Detect1	tws₁	17.9	22.4	26.9	ms	R _{RTW} = 10 kΩ
FAST/SLOW Normal Time1	toк1	13.6	15.7	17.8	ms	R _{RTW} = 10 kΩ
FAST Timeout Detect2	t _{WF2}	24.4	30.5	36.6	ms	R _{RTW} = 27 kΩ
SLOW Timeout Detect2	tws2	48.8	61.0	73.2	ms	R _{RTW} = 27 kΩ
FAST/SLOW Normal Time2	t _{ок2}	36.7	42.7	48.7	ms	R _{RTW} = 27 kΩ
FAST Timeout Detect3	t _{WF3}	42.5	53.2	63.8	ms	R _{RTW} = 47 kΩ
SLOW Timeout Detect3	tws3	85.0	106.3	127.6	ms	R _{RTW} = 47 kΩ
FAST/SLOW Normal Time3	tокз	63.9	74.4	84.9	ms	R _{RTW} = 47 kΩ
WDIN Detect Minimum Pulse Width	twdin	20	-	-	μs	
WDIN Initial Mask Time	t _{WDIM}	325	500	675	ms	
WDIN Pull-down Resistor Value	RWDIN	50	100	150	kΩ	V _{DD} < 5 V
WDIN Low Level Input Voltage	Vwdinl	-	-	0.2 x V _{DD}	V	
WDIN High Level Input Voltage	Vwdinh	0.8 x V _{DD}	-	-	V	
WDEN Pull-down Resistor Value	RWDEN	50	100	150	kΩ	$V_{DD} < 5 V$
WDEN Low Level Input Voltage	VWDENL	-	-	0.2 x V _{DD}	V	
WDEN High Level Input Voltage	Vwdenh	0.8 x V _{DD}	-	-	V	
WDOUT Output High Voltage	Vонwdo	V _{DD} -0.3	-	-	V	Iwdout = -3 mA
WDOUT Output Low Voltage	Volwdo	-	-	0.3	V	Iwdout = +3 mA

Typical Performance Curves

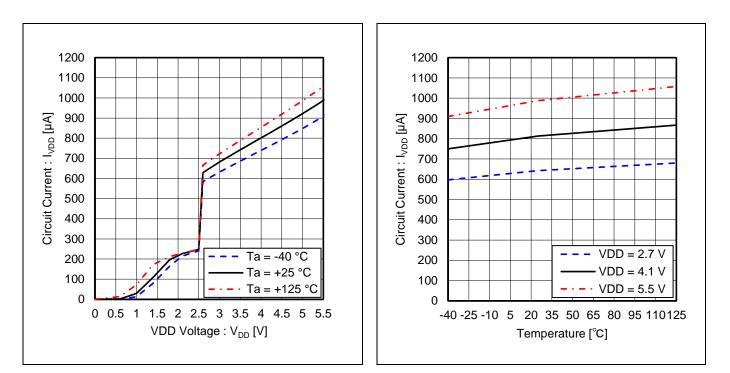


Figure 6. Circuit Current vs VDD Voltage

vs Temperature

("RSTIN UVD", V_{DD} = 3.3 V)

Figure 7. Circuit Current vs Temperature

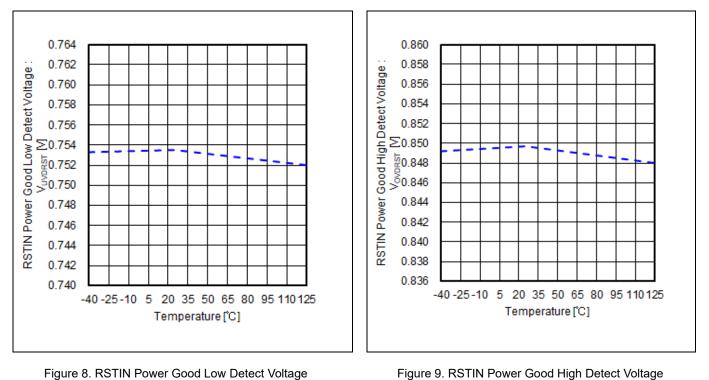
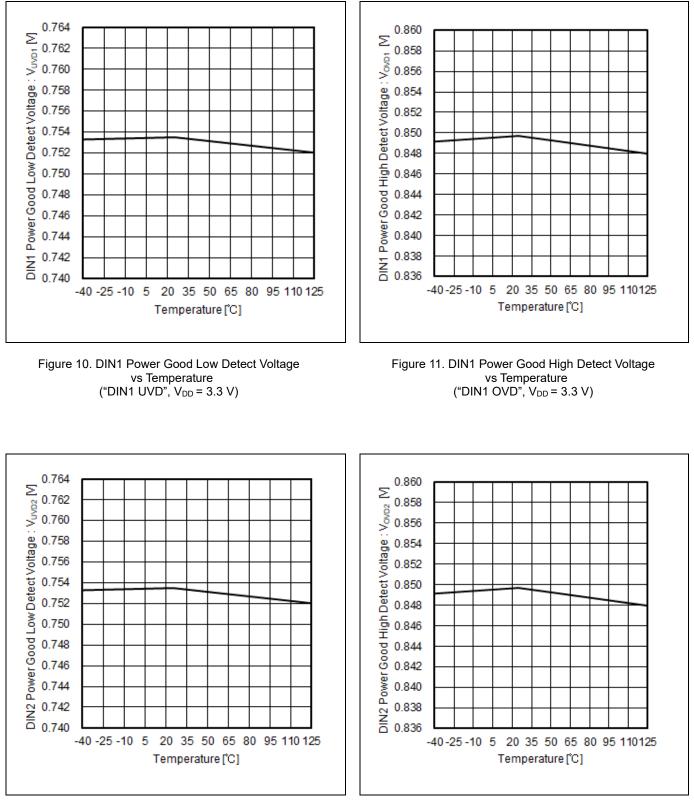


Figure 9. RSTIN Power Good High Detect Voltage vs Temperature ("RSTIN OVD", V_{DD} = 3.3 V)

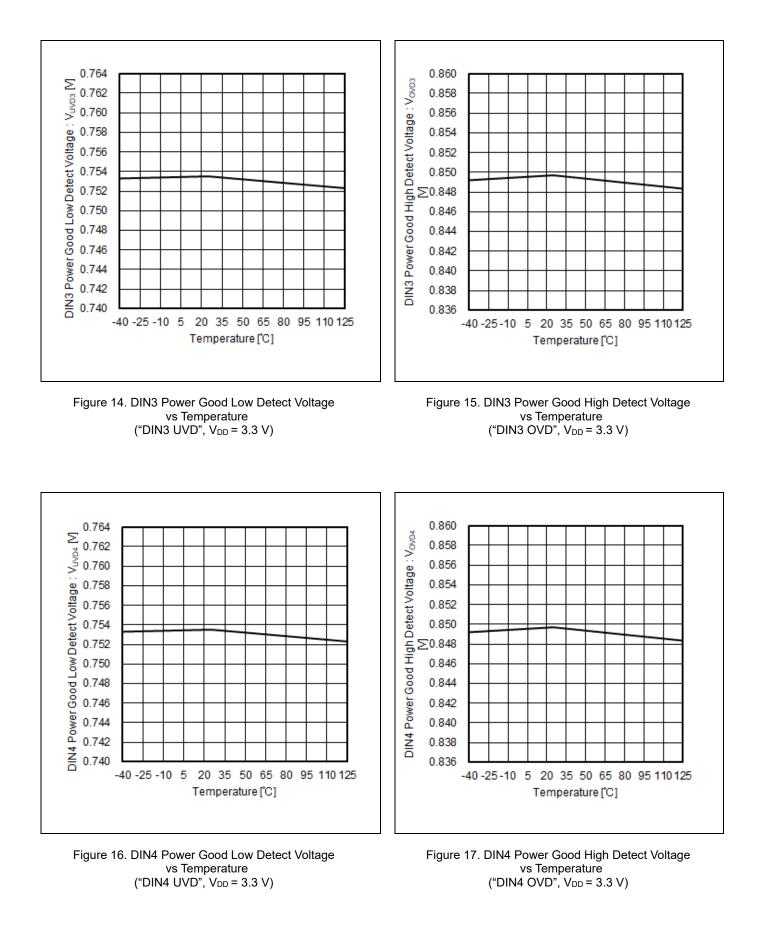
Typical Performance Curves - continued







Typical Performance Curves - continued



Typical Performance Curves - continued

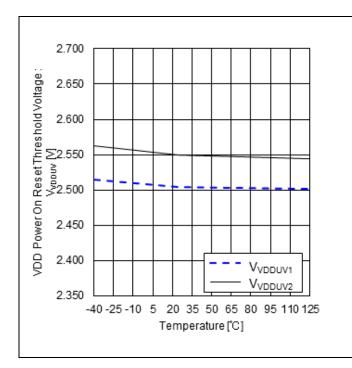


Figure 18. VDD Power On Reset Threshold Voltage vs Temperature

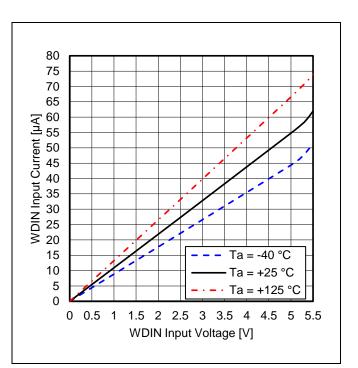
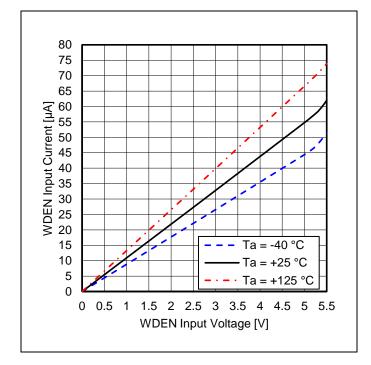


Figure 19. WDIN Input Current vs WDIN Input Voltage ("WDIN Pull-down Resistor Value", V_{DD} = 3.3 V)





Timing Chart

Figure 21 shows ON/OFF normal sequence.

When UVLO (2.55 V) monitoring the VDD pin is released, internal OSC for DIGITAL BLOCK starts to work. Then after RSTIN voltage reaches UVD release (adjustable), BIST starts to do self-test. If the BIST is normal, XRSTOUT goes to High at 10 ms after RSTIN UVD released. If the BIST is abnormal, XRSTOUT, PG1 to PG4 and WDOUT stays at Low.

XRSTOUT goes to Low when either of 2 monitoring functions (RSTIN UVD or Watchdog Timer) is detected.

WDT block has initial mask time (Typ 500 ms). Even though High voltage is given to WDEN within this time from RSTIN UVD released, WDT is not enabled. For example, as Figure 22 shows, in WDEN tied to VDD case WDT is enabled 500 ms after RSTIN UVD released.

WDEN input signal works as an enable of Watchdog Timer. Even though clocks are input to the WDIN pin, they are ignored as long as WDEN = Low. WDOUT is just a buffered version output of WDEN input and processor can use this output to confirm if WDEN is correctly controlled by itself.

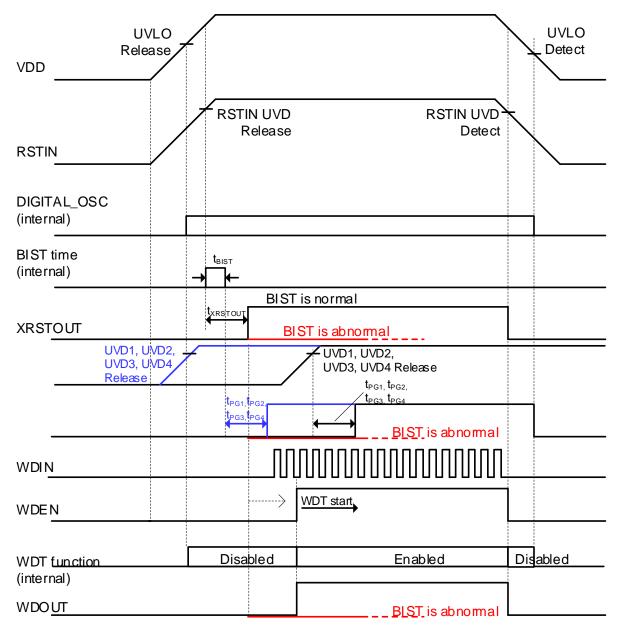


Figure 21. Power ON/OFF Normal Sequence (WDEN is controlled by external signal)

Timing Chart - continued

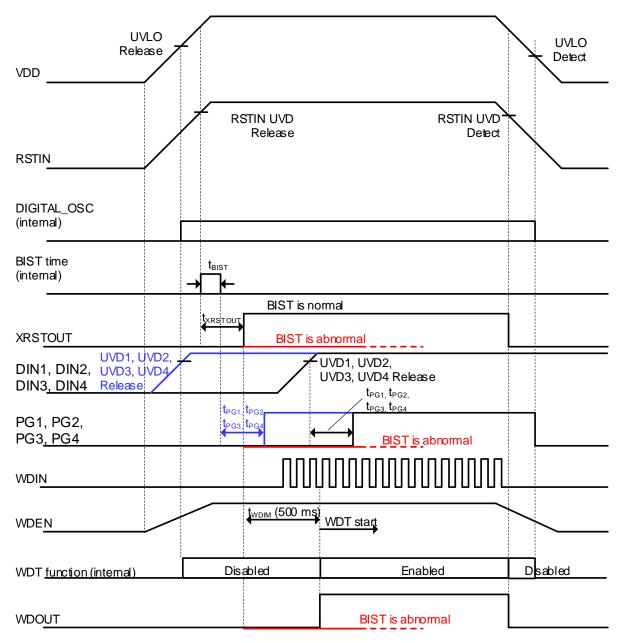


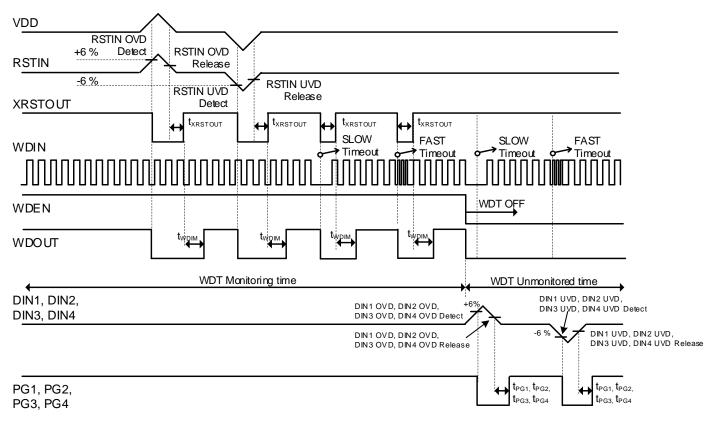
Figure 22. Power ON/OFF Sequence (WDEN is externally pulled-up to VDD)

Timing Chart - continued

Figure 23 shows monitoring sequence with VDD POR, DIN1, DIN2, DIN3 and DIN4 voltage detection and WDT monitoring. XRSTOUT is qualified by either of RSTIN (VDD POR) and WDIN (WDT) and outputs Low level voltage when these are detected.

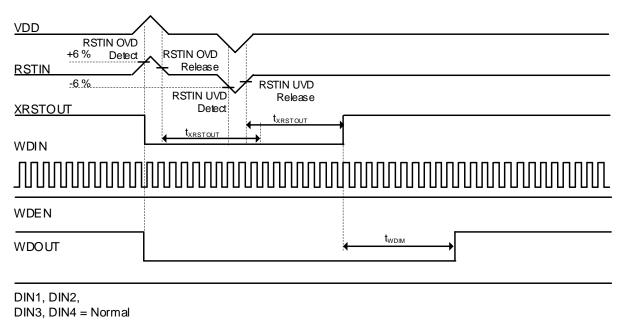
When WDEN becomes Low, WDOUT goes to Low and WDT stops to work immediately.

PG1, PG2, PG3 and PG4 are de-asserted immediately de-bounce time by internal DIGITAL de-bounce filter when the corresponding DIN1, DIN2, DIN3 and DIN4 input voltage goes out of PG window (±6 %). PG1, PG2, PG3 and PG4 are asserted at 10 ms after DIN1, DIN2, DIN3 and DIN4 input voltage becomes inside PG window.





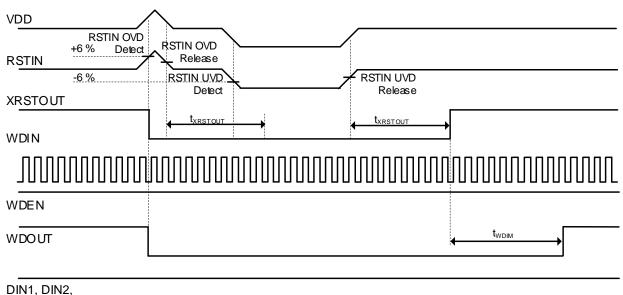
Timing Chart - continued



PG1, PG2,

PG3, PG4 = High

Figure 24. Monitoring Sequence 2 (2nd reset factor appears and disappears within t_{XRSTOUT})

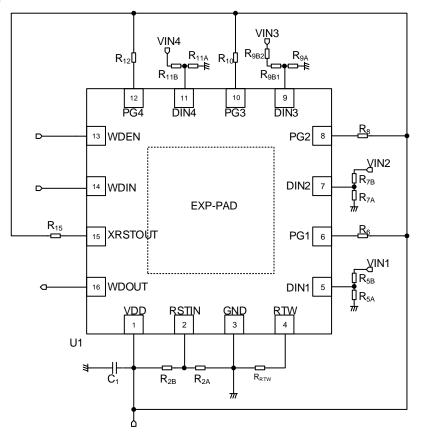


DIN3, DIN4 = Normal

PG1, PG2, PG3, PG4 = High

Figure 25. Monitoring Sequence 3 (2nd reset factor appears within txRSTOUT and disappear after txRSTOUT)

Application Example





Example of Constant Setting

V_{DD} = 3.3 V, R_{RTW} = 27 kΩ, VIN1 = 3.3 V, VIN2 = 1.8 V, VIN3 = 1.5 V, VIN4 = 1.2 V

			Range)			
Item	em Value Unit Parts Number Company				Min	Max	Unit
U1	-	-	BD39042MUF-C	ROHM	-	-	-
C1	0.22	μF	GCM188R71C224KA01	MURATA	0.1	1.0	μF
R _{2A}	2.4	kΩ	MCR01MZPD2401	ROHM	-	50	kΩ
R _{2B}	7.5	kΩ	MCR01MZPD7501	ROHM	-	50	kΩ
R _{RTW}	27	kΩ	MCR01MZPD2702	ROHM	10	47	kΩ
R _{5A}	2.4	kΩ	MCR01MZPD2401	ROHM	-	50	kΩ
R _{5B}	7.5	kΩ	MCR01MZPD7501	ROHM	-	50	kΩ
R ₆	10	kΩ	MCR01MZPD1002	ROHM	-	50	kΩ
R _{7A}	2.4	kΩ	MCR01MZPD2401	ROHM	-	50	kΩ
R _{7B}	3.0	kΩ	MCR01MZPD3001	ROHM	-	50	kΩ
R ₈	10	kΩ	MCR01MZPD1002	ROHM	-	50	kΩ
R _{9A}	5.6	kΩ	MCR01MZPD5601	ROHM	-	50	kΩ
R _{9B1}	3.6	kΩ	MCR01MZPD3601	ROHM	-	50	kΩ
R _{9B2}	1.3	kΩ	MCR01MZPD1301	ROHM	-	50	kΩ
R10	10	kΩ	MCR01MZPD1002	ROHM	-	50	kΩ
R 11A	1.8	kΩ	MCR01MZPD1801	ROHM	-	50	kΩ
R 11B	3.6	kΩ	MCR01MZPD3601	ROHM	-	50	kΩ
R ₁₂	10	kΩ	MCR01MZPD1002	ROHM	-	50	kΩ
R ₁₅	10	kΩ	MCR01MZPD1002	ROHM	-	50	kΩ

Application Example - continued

Procedure for Selecting Application Components

- (1) Selecting input capacitor (C1)
 - Place a ceramic capacitor connect to GND nearest the IC for the VDD pin.

D

Please consider temperature change, DC bias characteristics and dispersion enough. Also it is necessary to choose the product superior (over 0.1 μ F at least) in thermal characteristics such as B characteristics or X7R characteristics. The capacitor 1.5 times to 2 times larger than a limit is recommended about the pressure-resistant.

(2) Selecting a resistor of input pin (R_{2A}, R_{2B}, R_{5A}, R_{5B}, R_{7A}, R_{7B}, R_{9A}, R_{9B1}, R_{9B2}, R_{11A}, R_{11B}) Set a resistor divider voltage 0.8 V to input voltage V_{DD}, VIN1, VIN2, VIN3, VIN4. It is recommended to set the total input impedance to be within 10 k Ω . Because the detection voltage changes when noise is applied to the input pin.

$$RSTIN = 0.8 V = V_{DD} \times \frac{R_{2A}}{R_{2A} + R_{2B}}$$
$$DIN1 = 0.8 V = VIN1 \times \frac{R_{5A}}{R_{5A} + R_{5B}}$$
$$DIN2 = 0.8 V = VIN2 \times \frac{R_{7A}}{R_{7A} + R_{7B}}$$
$$IN3 = 0.8 V = VIN3 \times \frac{R_{9A}}{R_{9A} + R_{9B1} + R_{9B2}}$$
$$DIN4 = 0.8 V = VIN4 \times \frac{R_{11A}}{R_{11A}}$$

$$DIN4 = 0.8 V = VIN4 \times \frac{R_{11A}}{R_{11A} + R_{11B}}$$

 (3) Selecting a resistor of output pin (R₆, R₈, R₁₀, R₁₂, R₁₅) Set more than 10 kΩ pull-up resistor connects to the VDD pin. Please consider thermal characteristics and dispersion enough, and it is necessary to choose a resistor over 7.5 kΩ. The PG1 to PG4 pins can be connected and used to OR output. In this case also, choose total impedance over 7.5 kΩ.

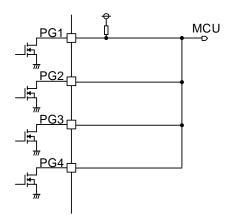


Figure 27. PG1, PG2, PG3, PG4 OR Output

(4) Selecting WDT (RRTW)

Place a resistor nearest GND for the RTW pin. Change RTW resistor value enables WDT setting time change. Settable range is 10 k Ω to 47 k Ω . Please consider thermal characteristics and dispersion enough, and it is necessary to choose component under ±1.0 %. Refer to Figure 5 for the setting time.

(5) EXP-PAD

The exposed thermal pad is highly recommended for GND connection.

Operational Notes

1. Reverse Connection of Power Supply

Connecting the power supply in reverse polarity can damage the IC. Take precautions against reverse polarity when connecting the power supply, such as mounting an external diode between the power supply and the IC's power supply pins.

2. Power Supply Lines

Design the PCB layout pattern to provide low impedance supply lines. Furthermore, connect a capacitor to ground at all power supply pins. Consider the effect of temperature and aging on the capacitance value when using electrolytic capacitors.

3. Ground Voltage

Ensure that no pins are at a voltage below that of the ground pin at any time, even during transient condition.

4. Ground Wiring Pattern

When using both small-signal and large-current ground traces, the two ground traces should be routed separately but connected to a single ground at the reference point of the application board to avoid fluctuations in the small-signal ground caused by large currents. Also ensure that the ground traces of external components do not cause variations on the ground voltage. The ground lines must be as short and thick as possible to reduce line impedance.

5. Recommended Operating Conditions

The function and operation of the IC are guaranteed within the range specified by the recommended operating conditions. The characteristic values are guaranteed only under the conditions of each item specified by the electrical characteristics.

6. Inrush Current

When power is first supplied to the IC, it is possible that the internal logic may be unstable and inrush current may flow instantaneously due to the internal powering sequence and delays, especially if the IC has more than one power supply. Therefore, give special consideration to power coupling capacitance, power wiring, width of ground wiring, and routing of connections.

7. Testing on Application Boards

When testing the IC on an application board, connecting a capacitor directly to a low-impedance output pin may subject the IC to stress. Always discharge capacitors completely after each process or step. The IC's power supply should always be turned off completely before connecting or removing it from the test setup during the inspection process. To prevent damage from static discharge, ground the IC during assembly and use similar precautions during transport and storage.

Operational Notes - continued

8. Inter-pin Short and Mounting Errors

Ensure that the direction and position are correct when mounting the IC on the PCB. Incorrect mounting may result in damaging the IC. Avoid nearby pins being shorted to each other especially to ground, power supply and output pin. Inter-pin shorts could be due to many reasons such as metal particles, water droplets (in very humid environment) and unintentional solder bridge deposited in between pins during assembly to name a few.

9. Unused Input Pins

Input pins of an IC are often connected to the gate of a MOS transistor. The gate has extremely high impedance and extremely low capacitance. If left unconnected, the electric field from the outside can easily charge it. The small charge acquired in this way is enough to produce a significant effect on the conduction through the transistor and cause unexpected operation of the IC. So unless otherwise specified, unused input pins should be connected to the power supply or ground line.

10. Regarding the Input Pin of the IC

This monolithic IC contains P+ isolation and P substrate layers between adjacent elements in order to keep them isolated. P-N junctions are formed at the intersection of the P layers with the N layers of other elements, creating a parasitic diode or transistor. For example (refer to figure below):

When GND > Pin A and GND > Pin B, the P-N junction operates as a parasitic diode. When GND > Pin B, the P-N junction operates as a parasitic transistor.

Parasitic diodes inevitably occur in the structure of the IC. The operation of parasitic diodes can result in mutual interference among circuits, operational faults, or physical damage. Therefore, conditions that cause these diodes to operate, such as applying a voltage lower than the GND voltage to an input pin (and thus to the P substrate) should be avoided.

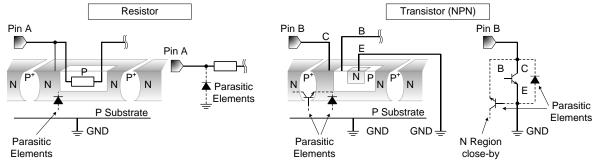


Figure 28. Example of Monolithic IC Structure

11. Ceramic Capacitor

When using a ceramic capacitor, determine a capacitance value considering the change of capacitance with temperature and the decrease in nominal capacitance due to DC bias and others.

12. Functional Safety

"ISO 26262 Process Compliant to Support ASIL-*"

A product that has been developed based on an ISO 26262 design process compliant to the ASIL level described in the datasheet.

"Safety Mechanism is Implemented to Support Functional Safety (ASIL-*)"

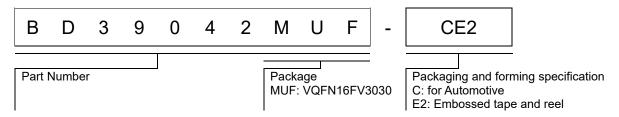
A product that has implemented safety mechanism to meet ASIL level requirements described in the datasheet.

"Functional Safety Supportive Automotive Products"

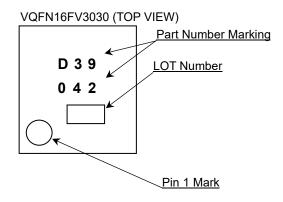
A product that has been developed for automotive use and is capable of supporting safety analysis with regard to the functional safety.

Note: "ASIL-*" is stands for the ratings of "ASIL-A", "-B", "-C" or "-D" specified by each product's datasheet.

Ordering Information



Marking Diagram



Datasheet

Physical Dimension and Packing Information VQFN16FV3030 Package Name $3. 0 \pm 0. 1$ -0 + 0. с. С Q 1 PIN MARK OMAX S £ ÷ 030 с Г 2 2)0 2 +0. ġ 0. 08S ė 0. $4\pm0.$ CO. 2 0. 5 U U 16--- $4\pm0.$ $4\pm0.$ 0 13 n \cap ſ 1|20. $25 \substack{+0.05 \\ -0.04}$ 0.75 (UNIT:mm) PKG:VQFN16FV3030 Drawing No. EX396-5001 NOTE: Dimensions in () for reference only. < Tape and Reel Information > Таре Embossed carrier tape 3000pcs Quantity Direction of feed E2 The direction is the pin 1 of product is at the upper left when you hold reel on the left hand and you pull out the tape on the right hand Ο 0 Ο Ο 0 Ο 0 Ο 0 Ο 0 Ο E2 TR E2 TR E2 Ì TR E2 TR E2 TR E2 TR ΤL E1 ΤL E1 ΤL E1 ΤL E1 ΤL E1 ΤL E1 Direction of feed Pocket Quadrants Reel

Revision History

Date	Revision	Changes
15.May.2021	Rev.001	New Release

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CLASSII	CLASSⅢ	CLASS II b	
CLASSⅣ	CLASSI	CLASSII	CLASSⅢ

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 - [e] Use of our Products in proximity to heat-producing components, plastic cords, or other flammable items
 - [f] Sealing or coating our Products with resin or other coating materials
 - [g] Use of our Products without cleaning residue of flux (Exclude cases where no-clean type fluxes is used. However, recommend sufficiently about the residue.); or Washing our Products by using water or water-soluble cleaning agents for cleaning residue after soldering
 - [h] Use of the Products in places subject to dew condensation
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 - [d] the Products are exposed to high Electrostatic
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