

# **Three Phase Brushless Motor Pre-driver**

# **BD63003MUV**

#### **General Description**

BD63003MUV is a Pre-driver IC for 3-phase brushless motor. It generates a driving signal from the Hall sensor and applies PWM in the input control signal for motor control. It has a built-in booster circuit which allows Nch-Nch MOS transistors to be used as the external power Transistor. It is compatible with 12 V or 24 V power supply and has various controls and built-in protection functions, making it useful for variety of purposes. Since the IC adopts a small package, it can also be used on small diameter motors.

#### Features

- Built in 120° Commutation Logic Circuit
- Driving with Nch-Nch MOS Transistors
- Built in Peak Current Control Function
- For Controller Input of 3.3 V and 5 V
- PWM Control Mode (lower arm switching)
- CW/CCW Function
- Short Brake Function
- FG Output (1FG / 3FG conversion)
- Built-in Protection Circuit for Current Limit (CL), Overheating (TSD), Under Voltage (UVLO), Over Voltage (OVLO), Motor Lock (MLP)

#### Applications

- OA Machines
- Other General Civil Equipment

# **Typical Application Circuit**

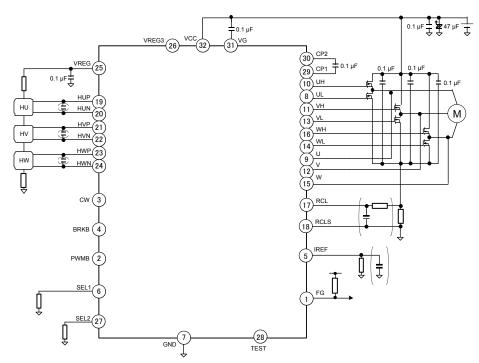
#### **Key Specifications**

Power Supply Voltage Rating:	40 V
Low Side Gate Drive Voltage:	10 V (Typ)
High Side Gate Drive Voltage:	10 V (Typ)
Operating Temperature Range:	-40 °C to +85 °C
Current Limit Detect Voltage:	0.2 V±10 %
UVLO Lockout Voltage:	6.0 V (Typ)
OVLO Lockout Voltage:	28.5 V (Typ)

#### Package VQFN032V5050

**W (Typ) x D (Typ) x H (Max)** 5.0 mm x 5.0 mm x 1.0 mm



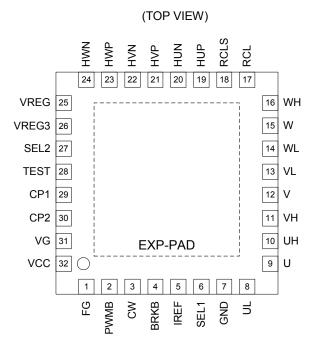


OProduct structure : Silicon integrated circuit OThis product has no designed protection against radioactive rays.

# Contents

.1
.1
.1
.1
.1
.1
.3
.3
.4
.5
11
11
11
12
14
15
17
19
19
20
21

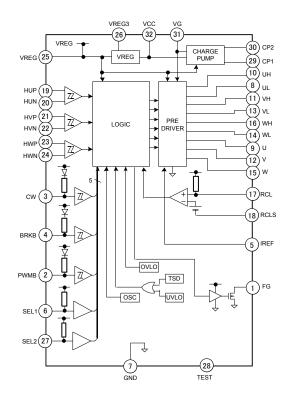
# **Pin Configuration**



# **Pin Descriptions**

Pin No.	Pin Name	Function	Pin No.	Pin Name	Function
1	FG	1FG / 3FG output	17	RCL	Detect voltage input for over current
2	PWMB	PWM input (negative logic)	18	RCLS	RCL sense input
3	CW	Changing direction of rotation (H: CW, L: CCW)	19	HUP	U phase hall input +
4	BRKB	Brake input (negative logic)	20	HUN	U phase hall input -
5	IREF	Output driving current setting	21	HVP	V phase hall input +
6	SEL1	Function setting input 1	22	HVN	V phase hall input -
7	GND	GND	23	HWP	W phase hall input +
8	UL	U phase lower output	24	HWN	W phase hall input -
9	U	U phase output feedback	25	VREG	VREG output
10	UH	U phase upper output	26	VREG3	VREG3 output
11	VH	V phase upper output	27	SEL2	Function setting input 2
12	V	V phase output feedback	28	TEST	TEST input (GND)
13	VL	V phase lower output	29	CP1	Charge pump setting 1
14	WL	W phase lower output	30	CP2	Charge pump setting 2
15	W	W phase output feedback	31	VG	Charge pump output
16	WH	W phase upper output	32	VCC	Power supply
-	EXP-PAD	Connect EXP-PAD to GND			

# **Block Diagram**



# **Description of Blocks**

Commutation logic (120° commutation)

	Truth Table										
HU	Η٧	НW		CW (CW = H d			r OPEN)			FG	
по	пv		UH	UL	VH	VL	WH	WL	1FG	3FG	
Н	L	Н	PWM	PWM <sup>(Note 1)</sup>	Н	L	L	L	L	Hi-z	
н	L	L	PWM	PWM <sup>(Note 1)</sup>	L	L	Н	L	L	L	
Н	Н	L	L	L	PWM	PWM <sup>(Note 1)</sup>	Н	L	L	Hi-z	
L	Н	L	Н	L	PWM	PWM <sup>(Note 1)</sup>	L	L	Hi-z	L	
L	Н	Н	Н	L	L	L	PWM	PWM <sup>(Note 1)</sup>	Hi-z	Hi-z	
L	L	Н	L	L	Н	L	PWM	PWM <sup>(Note 1)</sup>	Hi-z	L	
Н	Н	Н	L	L	L	L	L	L	L	L	
L	L	L	L	L	L	L	L	L	Hi-z	Hi-z	

HU	HV	HW		CCW (CW = L)				FG		
по	ΠV	ΠVV	UH	UL	VH	VL	WH	WL	1FG	3FG
Н	L	Н	Н	L	PWM	PWM <sup>(Note 1)</sup>	L	L	L	Hi-z
Н	L	L	Н	L	L	L	PWM	PWM <sup>(Note 1)</sup>	L	L
Н	Н	L	L	L	Н	L	PWM	PWM <sup>(Note 1)</sup>	L	Hi-z
L	Н	L	PWM	PWM <sup>(Note 1)</sup>	Н	L	L	L	Hi-z	L
L	Н	Н	PWM	PWM <sup>(Note 1)</sup>	L	L	Н	L	Hi-z	Hi-z
L	L	Н	L	L	PWM	PWM <sup>(Note 1)</sup>	Н	L	Hi-z	L
Н	Н	Н	L	L	L	L	L	L	L	L
L	L	L	L	L	L	L	L	L	Hi-z	Hi-z

(Note 1) When PWM = "L", PWM = "H". When PWMB = H, PWMB = L.

Regulator Output Pin (VREG)
 This is constant voltage output pin of 5 V (Typ). Connect capacitors of 0.01 μF to 1 μF. Be careful that VREG current does not exceed absolute maximum ratings in case of being used for bias power supply of hall elements.

2. Regulator Output Pin (VREG3)

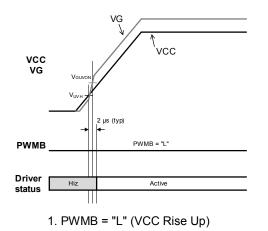
This is constant voltage output pin of 3.3 V (Typ). VREG3 can be used for bias voltage of hall elements. Be careful that VREG3 current does not exceed absolute maximum ratings in case of being used.

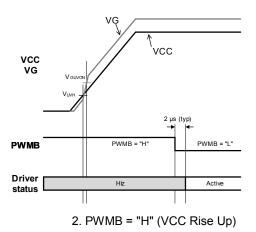
Figure 1. VREG3 reference circuit

3. PWM Input Pin (PWMB)

Speed can be controlled by inputting PWM signal into the PWMB pin (negative logic). Synchronous rectifier PWM can be achieved through lower switching. When PWMB = "L", lower external FET that matches to Hall input logic is "L". When PWMB = "H" or open, lower external FET is "H". When PWMB = "H" or OPEN status is detected for 104  $\mu$ s (Typ), the synchronous rectifier is OFF. Synchronous rectifier turns ON through the falling edges of subsequent PWMB signals. At startup, External FET keeps "Hi-z" states in which all phase is OFF (stand-by) until PWMB = "L" status is detected 2  $\mu$ s (Typ) or more [figure 2]. However, the internal regulator of VREG, VG, the other regulator, protection function of OVLO, and the other ones are operated in the stand-by. Additionally, the PWMB pin is pulled up by internal 3.3V (Typ) through a resistance of 100 k $\Omega \pm 300$  k $\Omega$ .

PWMB	PWM phase Lower External FET
H or OPEN	OFF
L	ON







4. BRKB Pin (BRKB)

Motor rotation can be quickly stopped using the BRKB Pin (negative logic). When BRKB = "L", this causes all the upper external FET to turn "OFF" and all the lower external FET to turn "ON", initiating short break operation. When BRKB = "H" or OPEN, then short brake operation will be released. Additionally, the BRKB pin is pulled up by internal 3.3V (Typ) through a resistance of 100 k $\Omega \pm 30$  k $\Omega$  and pulled down by GND through a resistance of 1000 k $\Omega \pm 300$  k $\Omega$ .

BRKB	Operation
H or OPEN	Normal
L	Short brake

5. Rotatory Direction Change Pin (CW)

Rotation direction can be switched with the CW pin. When CW = "H" or OPEN, the direction is Clockwise (CW). When CW = "L", the direction will be Counter Clockwise (CCW). We do not recommend switching rotation direction when motor is rotating. If rotation direction is switched when rotating, the operation is the following due to the condition of SEL1.

(1) SBRK = "Enable" (SEL1 = "H" or "M1")

After having performed short brakes movement until hall frequency becomes approximately 40 Hz (Typ) or less, rotatory direction is replaced. In the case of this condition, do not change the logic of CW for 10 ms after brakes cancellation by the BRKB input (Figure 3).

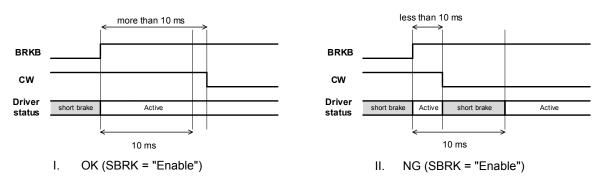


Figure 3. BRKB, CW Control Timing Limitation 1

(2) SBRK = "Disable" (SEL1 = "M2" or "L")

Without the short brake such as (1), direction is replaced. In this case, be careful since high current may sometimes flow in the external FET when the direction is replaced as described. In addition, there is no limitation in timing of BRKB and CW such as (1) in this condition (Figure 4).

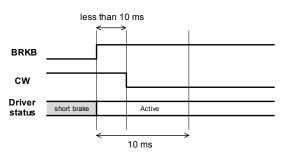




Figure 4. BRKB, CW Control Timing Limitation 2

In addition, the CW pin is pulled up by internal 3.3 V (Typ) through resistance of 100 k $\Omega$  ± 30 k $\Omega$  and pulled down by GND through a resistance of 1000 k $\Omega$  ± 300 k $\Omega$ .

CW	Direction
H or OPEN	CW
L	CCW

6. Select Pin 1 (SEL1)

The SEL1 pin can be used to select 1FG / 3FG and "Enable" or "Disable" of the short brakes (SBRK) at the direction change with the CW pin. The SEL1 pin is a 4 input pin and can be set by being OPEN, 220 k $\Omega$  [± 5 %] to GND, 47 k $\Omega$  [± 5 %] to GND, and tied to GND. In addition, the SEL1 pin is pulled up by VREG through resistance of 100 k $\Omega$  ± 30 k $\Omega$ .

SEL1	FG	SBRK(CW/CCW)
H (OPEN)	3FG	Enable
M1 (220 kΩ [±5 %] to GND)	1FG	Enable
M2 (47 kΩ [±5 %] to GND)	1FG	Disable
L (tied to GND)	3FG	Disable

#### 7. Select Pin 2 (SEL2)

The SEL2 pin can be used to select Enable/Disable of MLP and OVLO. In addition, the SEL2 pin is pulled up by VREG through resistance of 100 k $\Omega$  ± 30 k $\Omega$ .

SEL2	MLP	OVLO
H (OPEN)	2.2 s (Typ)	28.5 V (Typ)
M1 (220 kΩ [±5 %] to GND)	Disable	28.5 V (Typ)
M2 (47 kΩ [±5 %] to GND)	2.2 s (Typ)	Disable
L (tied to GND)	Disable	Disable

#### 8. FG Output pin (FG)

FG signal is composed by a hall signal and is outputted by the FG pin. Changing between 1FG and 3FG can be done with SEL1. In addition, the FG pin, which is an open drain output, should be externally pulled-up by a resistance of the about 10 k $\Omega$  to 100 k $\Omega$ .

#### 9. Hall Input (Hall: HUP, HUN, HVP, HVN, HWP, HWN)

Hall comparator inside the IC is designed with hysteresis ( $\pm$ 15 mV (Typ)) in order to prevent malfunction due to noise. Always set correct bias current for the Hall element so that the amplitude of Hall input voltage will be the minimum input voltage (V<sub>HALLMIN</sub>) or more. Furthermore, the output of the comparator has a digital filter of 2 µs (Typ). However, if it can't prevent the noise, it is recommended to connect a ceramic capacitor with about 100 pF to 0.01 µF value between the input pins of the Hall comparator. The in-phase input voltage range (V<sub>HALLCM</sub>: 0 V to V<sub>REG</sub>-1.7 V (Typ)) is designed for Hall comparator, set within this range when applying bias to the Hall element. When all Hall inputs become "H" or "L", all external FETs will be "OFF" by the hall input abnormal detection circuit.

10. Booster Circuit

There is built-in booster circuit used to drive upper Nch MOS transistor. The VG pins can produce a boost voltage (the VCC voltage+10 V (Typ) through connecting capacitors between CP1 - CP2 and between VG - VCC. We recommend connected capacitors to be 0.1  $\mu$ F or more. Because CP1 and CP2 are oscillated, the capacitors have to be located near IC. If need, add GND line for the shield. In addition, because VG voltage are boosted from the voltage that based on VCC, if VCC voltage is instability, it can be caused the malfunction such as VG voltage rise up. Therefore, add capacitor between VCC and GND as necessary to stabilize VCC voltage when using large current and motor with large BEMF. Because there is built-in protection circuit for insufficient booster, when VG voltage is V<sub>GUVON</sub> (V<sub>CC</sub>+7 V (Typ)) or less, all external FETs will be "OFF".

11. Current Limit Circuit (CL Circuit)

Output current limit (Current Limit: CL) circuit can be formed by connecting a low resistance used for detecting current between the RCL pin and the RCLS pin. When RCL voltage is detected 0.2 V (Typ) or more, all lower external FET will be "OFF". It returns by itself after a set amount of time (32  $\mu$ s (Typ)). This operation does not synchronize with PWM signal input into the PWMB pin. In addition, in order to avoid misdetection of output current due to RCL noise, the IC sets up a noise-masking period (1  $\mu$ s to 2  $\mu$ s (Typ)). During the noise-masking period, current detection is disabled. RCLS is the sense line of RCL. If RCLS becomes OPEN, Current Limit may not be normally Function. Connect the RCLS pin to the GND nearest to the current sense resistor's pin.

12. Thermal Shutdown Circuit (TSD Circuit)

When chip temperature of driver IC rises and exceeds the set temperature (165 °C (Typ)), the thermal Shutdown circuit (Thermal Shutdown: TSD) begins working. At this time, all external FETs will be "OFF". In addition, the TSD circuit is designed with hysteresis (25 °C (Typ)), and will return to normal working condition when the chip temperature drops. Moreover, the purpose of the TSD circuit is to protect the driver IC from thermal breakdown, therefore, temperature of this circuit will be over working temperature when this circuit operates. Thus, thermal design should have sufficient margin, so do not take continuous use and operation of the circuit as precondition.

13. Under Voltage Lock Out Circuit (UVLO Circuit)

There is a built-in under voltage lockout circuit (Under Voltage Lockout: UVLO circuit) used to ensure the minimum power supply voltage for drive IC to work and to prevent error in the operation of IC. When VCC voltage declined to  $V_{UVL}$  (6 V (Typ)), all external FETs should be "OFF". At the same time, UVLO circuit is designed with hysteresis (1 V (Typ)), so when VCC voltage reaches  $V_{UVH}$  (7 V (Typ)) or more, it will enter normal operation.

14. Over Voltage Lock Out Circuit (OVLO Circuit)

There is built-in over voltage lockout circuit (Over Voltage Lockout: OVLO circuit) used to restrain the increase of VCC voltage when motor is decelerating. When VCC voltage is 28.5 V (Typ) or more, short brake action will be conducted. In order to avoid misdetection, the IC sets up a noise-masking period (2 µs to 3 µs (Typ)). The short brake operation is released after a certain period of time (4 ms) when the VCC voltage is less than or equal to 27.5 V (Typ) and returns to normal operation. OVLO function does not work in case of SEL2 = "Disable"

15. Motor Lock Protection Circuit (MLP Circuit)

There is built-in motor lock protection circuit (Motor Lock Protection: MLP), ON/OFF of MLP circuit can be set by the SEL2 pin. When the MLP setting of SEL2 = "Enable" and the Hall signal logic does not change for 2.2 s (Typ) or more, all external FETs will all be latched as "OFF". Latch can be released through switching BRKB/CW logic. Moreover, when PWMB = "H" or OPEN state is detected for about 15 ms (Typ), latch can be released by the falling edges of subsequent PWMB. However, the MLP circuit does not operate when the MLP setting of the SEL2 pin is "Disable" and when the short brake (including when switching the direction of rotation) or the TSD circuit is in operation.

16. Predriver Output

The drive signal generated by the internal logic outputs the drive signal to the external output power transistor. Driving voltage of upper gate is VG voltage ( $V_{CC}$ +10 V (Typ)) and driving voltage of lower gate is the internal REG voltage (10 V (Typ)). In addition, a dead time (0.2  $\mu$ s (Typ)) is designed between the driving signals of upper gate and lower gate in order to prevent the upper and lower FET from being set to ON at the same time during synchronous rectifier PWM operation. Due to the influence of the motor's counter electromotive force, the output feedback pin (U, V, W) might swing under GND potential, which can cause malfunctions or destruction. When negative potential exceeds -2 V (min), Schottky diode can be inserted to prevent malfunction or destruction.

17. Pre-driver Output Peak Current Setting Pin (IREF)

A current of the pre-driver output can be set by connecting a resistor between the IREF Pin and GND. Note that if the IREF pin is connected with GND or open, it may cause malfunction. The range of the resistance is 27 k $\Omega$  [±5 %] to 150 k $\Omega$  [±5 %]. About the approximate value of the output current, refer to the following table.

Resistor Value [kΩ]	Output Source Current [mA]	Output Sink Current [mA]
150	16	27
120	18	33
100	22	40
82	26	48
68	31	58
56	36	68
47	42	84
39	48	96
33	55	113
27	63	136

Figure 5. The reference value of Pre-driver output current

18. Control Signal Sequence

Though we recommend you to input control signals of the CW, PWMB, and BRKB pins after inputting VCC, there won't be any problem if done otherwise. However, if MLP = "Enable" is set at startup, the MLP circuit will not be able to start the motor if the rotation of the motor is not detected within the set time (the edge of the FG signal is not input). Moreover, the control signal and the IC internal signal are given priority. Refer to the table below.

Priority of Control Signal				
Priority	Input / Internal signals			
1st	UVLO			
2nd	BRKB↑↓ <sup>(Note 2)</sup> , CW↑↓ <sup>(Note 2)</sup> , PWMB↓ <sup>(Note 2)</sup>			
3rd	TSD, MLP, HALLERR			
4th	OVLO			
5th	VG_UVLO, stand-by			
6th	BRKB			
7th	CL			
8th	PWMB, CW,			

(*Note 2*) ↑↓ means rising and falling edge of the signal. Refer to a figure of in condition transition for the signal name.

# Absolute Maximum Ratings (Ta = 25 °C)

Parameter	Symbol	Rating	Unit
Power Supply Voltage	Vcc	-0.3 to +40.0	V
VG Voltage	V <sub>G</sub>	-0.3 to +52.0	V
External FET Output Feedback Voltage	V(U, V, W)	40	V
FG Voltage	V <sub>FG</sub>	-0.3 to +7.0	V
RCL Voltage	V <sub>RCL</sub>	-0.3 to +5.5	V
RCLS Voltage	V <sub>RCLS</sub>	-0.3 to +0.3	V
Voltage of Input of Control and Hall Pins	V <sub>I/O</sub>	-0.3 to +7.0	V
FG Output Current	lfg	5	mA
VREG Output Current	I <sub>VREG</sub>	-30	mA
VREG3 Output Current	Ivreg3	-5	mA
Maximum Junction Temperature	Tjmax	150	°C
Storage Temperature Range	Tstg	-55 to +150	°C

Caution 1: Operating the IC over the absolute maximum ratings may damage the IC. The damage can either be a short circuit between pins or an open circuit between pins and the internal circuitry. Therefore, it is important to consider circuit protection measures, such as adding a fuse, in case the IC is operated over the absolute maximum ratings.

Caution 2: Should by any chance the maximum junction temperature rating be exceeded the rise in temperature of the chip may result in deterioration of the properties of the chip. In case of exceeding this absolute maximum rating, design a PCB with thermal resistance taken into consideration by increasing board size and copper area so as not to exceed the maximum junction temperature rating.

#### Thermal Resistance (Note 3)

Parameter	Symbol	Thermal Re	Linit	
Parameter	Symbol	1s <sup>(Note 5)</sup>	2s2p <sup>(Note 6)</sup>	- Unit
VQFN032V5050				
Junction to Ambient	θја	138.9	39.1	°C/W
Junction to Top Characterization Parameter <sup>(Note 4)</sup>	$\Psi_{JT}$	11	5	°C/W
(Note 3) Based on JESD51-2A (Still-Air).				

(Note 4) The thermal characterization parameter to report the difference between junction temperature and the temperature at the top center of the outside surface of the component package.

(Note 5) Using a PCB board based on JESD51-3. (Note 6) Using a PCB board based on JESD51-5. 7

(Note 6) Using a PCB board based	011 JE 0 D J 1= 5, 7	•				
Layer Number of Measurement Board	Material	Board Size				
Single	FR-4	114.3 mm x 76.2 mm x	1.57 mmt			
Тор						
Copper Pattern	Thickness					
Footprints and Traces	70 µm					
Layer Number of	Material	Board Size		Thermal \		
Measurement Board	Material	Board Oize		Pitch		Diameter
4 Layers	FR-4	114.3 mm x 76.2 mm	v 1.6 mmt	1.20 mm	ወ	0.30 mm
+ Edyclo	I	114.5 1111 × 70.2 1111	x 1.0 mm	1.20 11111	Ŧ	0.00 11111
Тор		2 Internal Laye		Botte	-	0.00 mm
,	Thickness				om	Thickness

(Note 7) This thermal via connects with the copper pattern of all layers.

#### **Recommended Operating Conditions**

Parameter	Symbol	Min	Тур	Max	Unit
Operating Temperature	Topr	-40	+25	+85	°C
Power Supply Voltage	Vcc	10.8	24.0	26.4	V

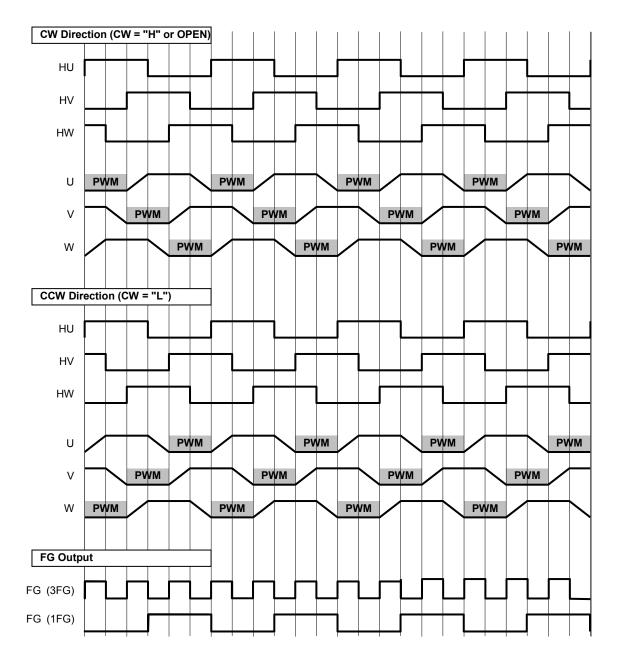
# Electrical Characteristics (Unless otherwise specified V<sub>cc</sub> = 24 V, Ta = 25 °C)

ctrical Characteristics (Unles	ss otherwise	specified	VCC - 24	v, ia – 25	<u>()</u>	
Parameter	Symbol	Min	Тур	Max	Unit	Conditions
Whole					1	
Circuit Electric Current	lcc	-	4.5	9	mA	
The VREG Voltage	V <sub>REG</sub>	4.5	5.0	5.5	V	I <sub>VREG</sub> = -10 mA
The VREG3 Voltage	V <sub>REG3</sub>	3.0	3.3	3.6	V	I <sub>VREG3</sub> = -1 mA
Booster Circuit						
VG Voltage	V <sub>G1</sub>	Vcc+7	Vcc+10	Vcc+11.5	V	I <sub>VG</sub> = 15 mA
VG UVLO Voltage	VGUVON	Vcc+6	Vcc+7	Vcc+8	V	
Driver Output					1	
High Side V <sub>GS</sub> Gate Drive Voltage	V <sub>GSH1</sub>	7	10	11.5	V	
Low Side V <sub>GS</sub> Gate Drive Voltage	V <sub>GSL1</sub>	7	10	12.5	V	
Source Electric Current Setting Range	IRNG1	-	-	70	mA	R <sub>IREF</sub> = 27 kΩ
Sink Electric Current Setting Range	IRNG2	-	-	140	mA	R <sub>IREF</sub> = 27 kΩ
Output Peak Source Current	Іон	-	18	-	mA	R <sub>IREF</sub> = 120 kΩ
Output Peak Sink Current	I <sub>OL</sub>	-	33	-	mA	R <sub>IREF</sub> = 120 kΩ
IREF Voltage	VIREF		1.2		V	
Hall Input				1		
Input Bias Current	HALL	-2.0	-0.1	+2.0	μA	V <sub>IN</sub> = 0 V
Range of In-phase Input Voltage	VHALLCM	0	-	V <sub>REG</sub> -1.7	V	
Minimum Input Voltage	VHALLMIN	50	-	-	mVp-p	
HYS Level +	VHALLHY+	5	15	25	mV	
HYS Level -	VHALLHY-	-25	-15	-5	mV	
Input of Control: PWMB, CW, B	RKB					
Input Electric Current	l <sub>iN</sub>	-46	-33	-20	μA	$V_{IN} = 0 V$
The Input H Voltage	VINH	2.0	-	5.5	V	
The Input L Voltage	V <sub>INL</sub>	0	-	0.8	V	
The Smallest Input Pulse Width	<b>t</b> PLSMIN	1	-	-	ms	CW, BRKB
Input Frequency Range	fрwм	10	-	50	kHz	PWMB
Input of Control: SEL1, SEL2						
Input Current	ISEL	-80	-50	-30	μA	V <sub>SEL</sub> = 0 V
FG Output						
Output Voltage L	VFGOL	0	0.1	0.3	V	I <sub>FG</sub> = 2 mA
Leak Current	IFGLEAK	-	0	1	μA	V <sub>FG</sub> = 5 V
Current Limit						
The Detection Voltage	Vcl	0.18	0.20	0.22	V	
Input Bias Electric Current	IRCL	-32	-20	-12	μA	V <sub>RCL</sub> = 0V
Input Voltage Range	V <sub>RCL</sub>	-0.3	-	+1.0	V	

# Electrical Characteristics - continued (Unless otherwise specified V<sub>cc</sub> = 24 V, Ta = 25 °C)

Parameter	Symbol	Min	Тур	Max	Unit	Conditions
UVLO					Ľ	
Release Voltage	Vuvh	6.5	7.0	7.5	V	
Lockout Voltage	Vuvl	5.5	6.0	6.5	V	
OVLO						
Release Voltage	Vovl	26.5	27.5	28.5	V	OVLO Enable
Lockout Voltage	Vovh	27.5	28.5	29.5	V	OVLO Enable
MLP						
Motor Lock Protection Detect Time	t <sub>MLP</sub>	1.54	2.20	2.86	S	MLP Enable

# **Timing Chart**



# I/O Equivalence Circuits

Pin No	Pin Name	Equivalence Circuit	Pin No	Pin Name	Equivalence Circuit
1	FG	FG FG M M M	2 3 4	PWM B CW BRKB	VREG 100 kΩ 100 kΩ 10 kΩ
5	IREF	50 Ω	6 27	SEL1 SEL2	
8 13 14	UL VL WL		9 10 11 12 15 16	U UH VH V WH	
17 18	RCL RCLS	RCL     2 kΩ       1 25 kΩ       1 25 kΩ	19 20 21 22 23 24	HUP HUN HVP HVN HWP HWN	HUP HUN HVP HVN HWP HWN

# I/O Equivalence Circuits - continued

25	VREG	VCC VCC VREG 145 kΩ 50 kΩ	26	VREG3	VREG VREG3 64 kΩ 36 kΩ
28	TEST		29	CP1	
30 31 32	CP2 VG VCC				

# **Operational Notes**

#### 1. Reverse Connection of Power Supply

Connecting the power supply in reverse polarity can damage the IC. Take precautions against reverse polarity when connecting the power supply, such as mounting an external diode between the power supply and the IC's power supply pins.

#### 2. Power Supply Lines

Design the PCB layout pattern to provide low impedance supply lines. Furthermore, connect a capacitor to ground at all power supply pins. Consider the effect of temperature and aging on the capacitance value when using electrolytic capacitors.

#### 3. Ground Voltage

Ensure that no pins are at a voltage below that of the ground pin at any time, even during transient condition.

#### 4. Ground Wiring Pattern

When using both small-signal and large-current ground traces, the two ground traces should be routed separately but connected to a single ground at the reference point of the application board to avoid fluctuations in the small-signal ground caused by large currents. Also ensure that the ground traces of external components do not cause variations on the ground voltage. The ground lines must be as short and thick as possible to reduce line impedance.

#### 5. Recommended Operating Conditions

The function and operation of the IC are guaranteed within the range specified by the recommended operating conditions. The characteristic values are guaranteed only under the conditions of each item specified by the electrical characteristics.

#### 6. Inrush Current

When power is first supplied to the IC, it is possible that the internal logic may be unstable and inrush current may flow instantaneously due to the internal powering sequence and delays, especially if the IC has more than one power supply. Therefore, give special consideration to power coupling capacitance, power wiring, width of ground wiring, and routing of connections.

#### 7. Testing on Application Boards

When testing the IC on an application board, connecting a capacitor directly to a low-impedance output pin may subject the IC to stress. Always discharge capacitors completely after each process or step. The IC's power supply should always be turned OFF completely before connecting or removing it from the test setup during the inspection process. To prevent damage from static discharge, ground the IC during assembly and use similar precautions during transport and storage.

# **Operational Notes – continued**

#### 8. Inter-pin Short and Mounting Errors

Ensure that the direction and position are correct when mounting the IC on the PCB. Incorrect mounting may result in damaging the IC. Avoid nearby pins being shorted to each other especially to ground, power supply and output pin. Inter-pin shorts could be due to many reasons such as metal particles, water droplets (in very humid environment) and unintentional solder bridge deposited in between pins during assembly to name a few.

#### 9. Unused Input Pins

Input pins of an IC are often connected to the gate of a MOS transistor. The gate has extremely high impedance and extremely low capacitance. If left unconnected, the electric field from the outside can easily charge it. The small charge acquired in this way is enough to produce a significant effect on the conduction through the transistor and cause unexpected operation of the IC. So unless otherwise specified, unused input pins should be connected to the power supply or ground line.

#### 10. Regarding the Input Pin of the IC

This monolithic IC contains P+ isolation and P substrate layers between adjacent elements in order to keep them isolated. P-N junctions are formed at the intersection of the P layers with the N layers of other elements, creating a parasitic diode or transistor. For example (refer to figure below):

When GND > Pin A and GND > Pin B, the P-N junction operates as a parasitic diode. When GND > Pin B, the P-N junction operates as a parasitic transistor.

Parasitic diodes inevitably occur in the structure of the IC. The operation of parasitic diodes can result in mutual interference among circuits, operational faults, or physical damage. Therefore, conditions that cause these diodes to operate, such as applying a voltage lower than the GND voltage to an input pin (and thus to the P substrate) should be avoided.

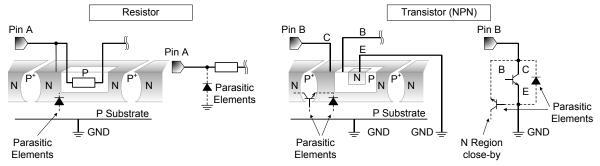


Figure 6. Example of Monolithic IC Structure

#### 11. Ceramic Capacitor

When using a ceramic capacitor, determine a capacitance value considering the change of capacitance with temperature and the decrease in nominal capacitance due to DC bias and others.

#### 12. Thermal Shutdown Circuit (TSD)

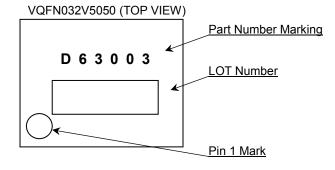
This IC has a built-in thermal shutdown circuit that prevents heat damage to the IC. Normal operation should always be within the IC's maximum junction temperature rating. If however the rating is exceeded for a continued period, the junction temperature (Tj) will rise which will activate the TSD circuit that will turn OFF power output pins. When the Tj falls below the TSD threshold, the circuits are automatically restored to normal operation.

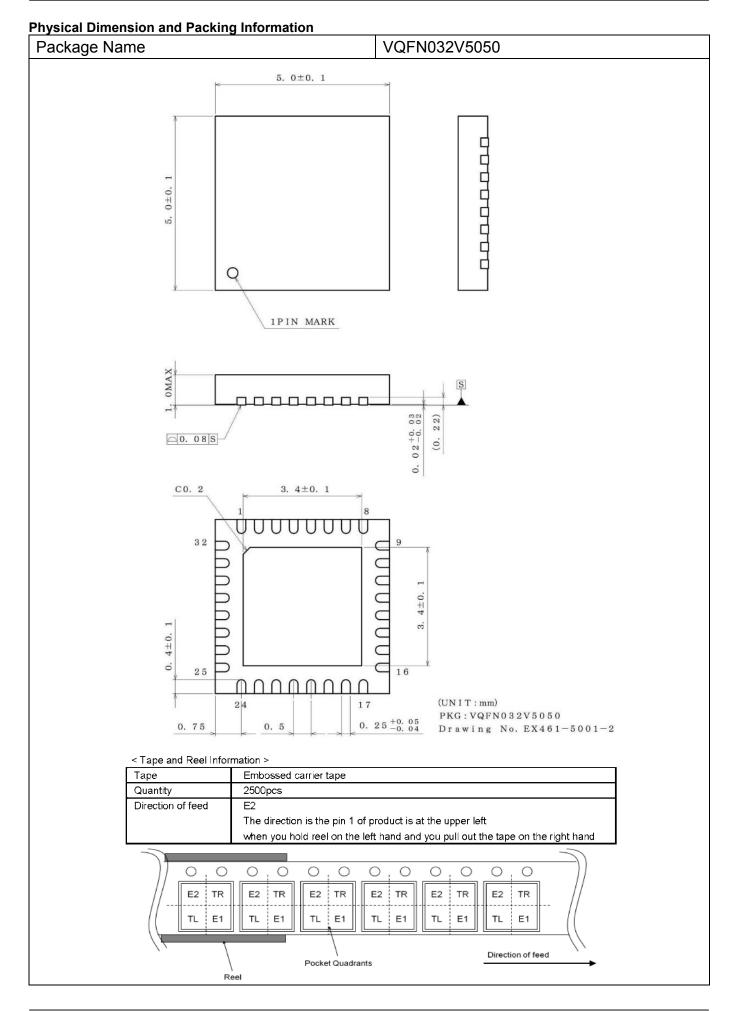
Note that the TSD circuit operates in a situation that exceeds the absolute maximum ratings and therefore, under no circumstances, should the TSD circuit be used in a set design or for any purpose other than protecting the IC from heat damage.

# **Ordering Information**



# **Marking Diagram**





# **Revision History**

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	Date	Revision	Changes
	10.Jul.2020	001	New Release

# Notice

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(Note1) Medical Equipment Classification of the Specific Applications
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JÁPAN	USA	EU	CHINA
CLASSⅢ	CLASSⅢ	CLASS II b	CLASSII
CLASSⅣ	CLASSIII	CLASSⅢ	CLASSI

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  - [h] Use of the Products in places subject to dew condensation
- 4. The Products are not subject to radiation-proof design.
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For details, please refer to ROHM Mounting specification

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  - [b] the temperature or humidity exceeds those recommended by ROHM
  - [c] the Products are exposed to direct sunshine or condensation
  - [d] the Products are exposed to high Electrostatic
- 2. Even under ROHM recommended storage condition, solderability of products out of recommended storage time period may be degraded. It is strongly recommended to confirm solderability before using Products of which storage time is exceeding the recommended storage time period.
- 3. Store / transport cartons in the correct direction, which is indicated on a carton with a symbol. Otherwise bent leads may occur due to excessive stress applied when dropping of a carton.
- 4. Use Products within the specified time after opening a humidity barrier bag. Baking is required before using Products of which storage time is exceeding the recommended storage time period.

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