

Sound Processor for Car Audio with Built-in High-Voltage Amplifier and 2nd Order Post Filter

BD37069FV-M

General Description

BD37069FV-M is a sound processor developed for car audio with built-in selector of six stereo inputs and output interfaced to ADC after adjusting signal level.

BD37069FV-M has a 6-channel volume circuit and built-in 2^{nd} order post filter which reduces the out-of-band noise. The High-Voltage function is capable to reach up to $5.2V_{RMS}$ maximum output. Furthermore, the IC is simple to design due to the built-in TDMA noise reduction systems.

Features

- AEC-Q100 Qualified (Note1)
- Built-in differential input selector that can select single-ended / differential input
- Reduce switching pop noise of input gain control due to the built-in advanced switch circuit
- Less out-of-band noise of DAC by built-in 2nd order post filter
- Built-in buffered ground isolation amplifier to achieve high CMRR characteristics
- Built-in TDMA noise reduction circuit reduces the additional components for external filter
- Available to output 5.2V_{RMS} by High-Voltage function (This device is possible to 3.2V_{RMS} output by using another High-Voltage mode, VCCH=11.5V)
- Available to control by 3.3V / 5V for I²C-bus controller
- The input and output terminals are located together to arrange the flow of signal in a same direction making the PCB layout easier and PCB area smaller

(Note 1) Grade 3

Key Specifications(Note2)

0.003%(Typ) ■Total Harmonic Distortion: ■Maximum Input Voltage: $2.1V_{RMS}(Typ)$ ■Common Mode Rejection Ratio : 55dB(Min) ■Maximum Output Voltage: $5.2V_{RMS}(Typ)$ ■Output Noise Voltage: $23\mu V_{RMS}(Typ)$ ■Residual Output Noise Voltage: $10.5\mu V_{RMS}(Typ)$ -70dB (Typ) ■Ripple Rejection Ratio: ■Operating Temperature Range: -40°C to +85°C (Note2)These specifications are High-Voltage mode2.

Package

SSOP-B40

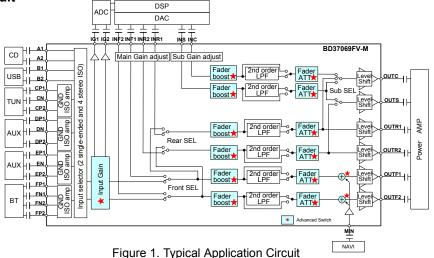
W(Typ) x D(Typ) x H(Max) 13.60mm x 7.80mm x 2.00mm



Applications

Car Audio and Other Audio Equipment





OProduct structure: Silicon monolithic integrated circuit OThis product has no designed protection against radioactive rays

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Pin Configuration

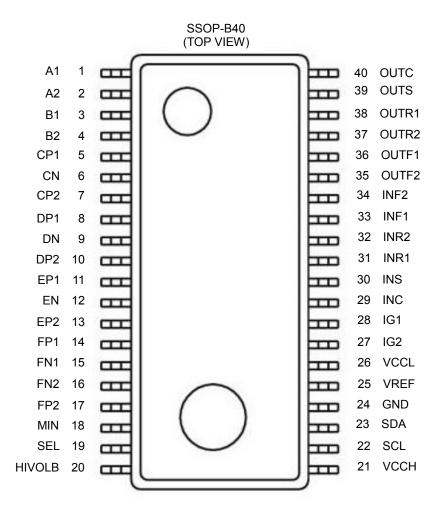


Figure 2.Pin Configuration

Pin Descriptions

n Descriptions										
Pin No.	Pin Name	Description	Pin No.	Pin Name	Description					
1	A1	A1 A input terminal of 1ch		VCCH	VCCH terminal for power supply					
2	A2	A input terminal of 2ch	22	SCL	I ² C-bus clock terminal					
3	B1	B input terminal of 1ch	23	SDA	I ² C-bus data terminal					
4	B2	B input terminal of 2ch	24	GND	GND terminal					
5	CP1	C positive input terminal of 1ch	25	VREF	BIAS terminal					
6	CN	C negative input terminal	26	VCCL	VCCL terminal for power supply					
7	CP2	C positive input terminal of 2ch	27	IG2	Input gain output terminal of 2ch					
8	DP1	D positive input terminal of 1ch	28	IG1	Input gain output terminal of 1ch					
9	DN	D negative input terminal	29	INC	Center input terminal					
10	DP2	D positive input terminal of 2ch	30	INS	Subwoofer input terminal					
11	EP1	E positive input terminal of 1ch	31	INR1	Rear input terminal of 1ch					
12	EN	E negative input terminal	32	INR2	Rear input terminal of 2ch					
13	EP2	E positive input terminal of 2ch	33	INF1	Front input terminal of 1ch					
14	FP1	F positive input terminal of 1ch	34	INF2	Front input terminal of 2ch					
15	FN1	F negative input terminal of 1ch	35	OUTF2	Front output terminal of 2ch					
16	FN2	F negative input terminal of 2ch	36	OUTF1	Front output terminal of 1ch					
17	FP2	F positive input terminal of 2ch	37	OUTR2	Rear output terminal of 2ch					
18	MIN	Mixing input terminal	38	OUTR1	Rear output terminal of 1ch					
19	SEL	High Voltage output mode Select	39	OUTS	Subwoofer output terminal					
20	HIVOLB	Output Gain control terminal	40	OUTC	Center output terminal					

Block Diagram

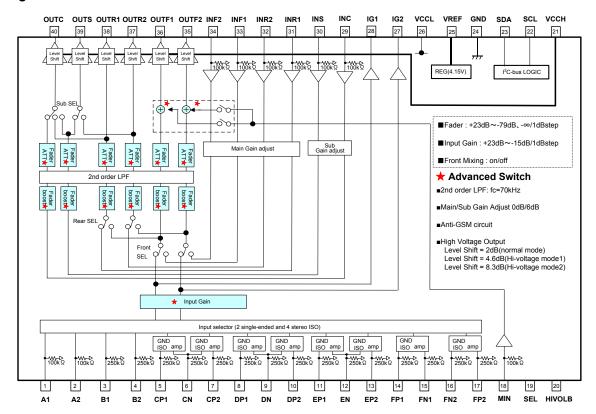


Figure 3. Block Diagram

- The outputs of Pin 27 and Pin 28 are selected by the input selector, from the inputs Pin 1 to Pin 17. Otherwise, these signals are possible to output directly on Pin 35 to Pin 40.
- 6-channel input signals from DSP on Pin 29 to Pin 34 pass through the volume circuit (Fader) and 2nd order post filter to the output terminals Pin 35 to Pin 40.
- It is possible for 6-channel inputs to set the gain up to +6dB by Gain adjust function and to set the gain up to +8.3dB by Level Shift Circuit (High-Voltage Mode).

Absolute Maximum Ratings (Ta=25°C)

Parameter	Symbol		Unit		
Dower Supply Voltage	V_{CCL}		10	V	
Power Supply Voltage	V_{CCH}		18		
Input Voltage	V	SCL, SDA	GND-0.3 to +7	1/	
Input Voltage	V _{IN}	Other	GND-0.3 to V _{CCL} +0.3	V	
Storage Temperature	T _{STG}		°C		
Maximum Junction Temperature	T_{JMAX}		°C		

Caution: Operating the IC over the absolute maximum ratings may damage the IC. The damage can either be a short circuit between pins or an open circuit between pins and the internal circuitry. Therefore, it is important to consider circuit protection measures, such as adding a fuse, in case the IC is operated over the absolute maximum ratings.

Thermal Resistance (Note 1)

Parameter		Thermal Res	Unit	
		1s ^(Note 3)	2s2p ^(Note 4)	Unit
SSOP-B40	•			
Junction to Ambient	θ_{JA}	103.6	58.8	°C/W
Junction to Top Characterization Parameter ^(Note 2)	Ψ_{JT}	17	10	°C/W

(Note 1)Based on JESD51-2A(Still-Air)

(Note 2)The thermal characterization parameter to report the difference between junction temperature and the temperature at the top center of the outside surface of the component package.

(Note 3)Using a PCB board based on JESD51-3.

(Note 4)Using a PCB board based on JESD51-7

Layer Number of Measurement Board	Material	Board Size
Single	FR-4	114.3mm x 76.2mm x 1.6mmt
Тор		
Copper Pattern	Thickness	
Footprints and Traces	70µm	

Layer Number of Measurement Board	Material	Board Size
4 Layers	FR-4	114.3mm x 76.2mm x 1.6mmt

Тор		2 Internal Laye	ers	Bottom			
Copper Pattern	Thickness	Copper Pattern	Thickness	Copper Pattern	Thickness		
Footprints and Traces	70µm	74.2mm x 74.2mm	35µm	74.2mm x 74.2mm	70µm		

Recommended Operating Condition (Ta= -40°C to +85°C)

Parameter	Symbol	Min	Тур	Max	Unit
Devices County Valtage	V _{CCL}	7.0	9	9.5	V
Power Supply Voltage	V_{CCH}	V_{CCL}	17	17.8	V

Electrical Characteristics

Unless otherwise specified, Ta=25°C, V_{CCL} =9V, V_{CCH} =17V, f=1kHz, V_{IN} =1 V_{RMS} , R_L =10k Ω , Input selector A, Input Gain 0dB, Gain Adjust +6dB, High-Voltage ON (High-Voltage mode2), LPF ON, Fader 0dB, Input point=A1/A2, Monitor point=IG1/IG2

				Limit			
Block	Parameter	Symbol	Min	Тур	Max	Unit	Conditions
General	Current Consumption (VCCL)	I _{Q_VCCL}	-	30	43	mA	No signal
Gel	Current Consumption (VCCH)	I _{Q_VCCH}	-	7	10	mA	No signal
	Input Impedance (A)	R _{IN_S}	70	100	130	kΩ	
	Input Impedance (B, C, D, E, F)	R _{IN_D}	175	250	325	kΩ	
	Voltage Gain	G _V	-1.5	0	1.5	dB	$G_V = 20log(V_{OUT}/V_{IN})$
	Channel Balance	СВ	-1.5	0	1.5	dB	$CB = G_{V1}-G_{V2}$
or	Total Harmonic Distortion	THD+N	-	0.003	0.05	%	$V_{OUT} = 1V_{RMS}$ BW = 400-30kHz
Input Selector	Output Noise Voltage (Note1)	V _{NO1}	-	3.1	8.0	μV_{RMS}	$R_G = 0\Omega$ BW = IHF-A
put S	Maximum Input Voltage	V _{IM}	2.0	2.2	-	V _{RMS}	V_{IM} at THD+N(V_{OUT}) = 1% BW = 400-30kHz
ıı	Crosstalk Between Channels (Note1)	СТС	-	-100	-90	dB	$R_G = 0\Omega$ CTC = 20log(V _{OUT} /V _{OUT} ') BW = IHF-A
	Crosstalk Between Selectors ^(Note1)	CTS	-	-100	-90	dB	$R_G = 0\Omega$ CTS = 20log(V _{OUT} /V _{OUT} ') BW = IHF-A
	Common Mode Rejection Ratio (C, D, E, F) (Note1)	CMRR	55	65	-	dB	XP1 and XN input XP2 and XN input CMRR = 20log(V _{IN} /V _{OUT}) BW = IHF-A, [X=C,D,E,F]
	Minimum Input Gain	G _{IN_MIN}	-17	-15	-13	dB	Input Gain = -15dB V_{IN} = 0.1 V_{RMS} G_{IN} = 20log(V_{OUT}/V_{IN})
Input Gain	Maximum Input Gain	G _{IN_MAX}	21	23	25	dB	Input Gain = 23dB V_{IN} = 0.1 V_{RMS} G_{IN} = 20log(V_{OUT}/V_{IN})
ndu	Gain Set Error	G _{IN_ERR}	-2	0	+2	dB	Input Gain = -15 to +23dB
	Output Impedance	Rout	-	-	50	Ω	
	Maximum Output Voltage	V _{OM}	2.0	2.2	-	V _{RMS}	THD+N = 1% BW = 400-30kHz

(Note1) VP-9690A (Average value detection, effective value display) filter by Panasonic is used for measurement. Input and output are in-phase.

Unless otherwise specified, Ta=25°C, $V_{\text{CCL}}=V_{\text{CCH}}=9V$, f=1kHz, $V_{\text{IN}}=0.9V_{\text{RMS}}$, $R_{\text{L}}=10\text{k}\Omega$, Input selector A, Input Gain 0dB, Gain Adjust +6dB, High-Voltage OFF(normal mode), LPF ON, Fader 0dB, Input point=INF1/INF2/INR1/INR2/INC/INS, Monitor point=OUTF1/OUTF2/OUTR1/OUTR2/OUTC/OUTS

~				Limit			
Block	Parameter	Symbol	Min	Тур	Max	Unit	Conditions
	Output Impedance	R _{OUT}	-	-	50	Ω	
el Shift	☆Maximum Output Voltage	V _{OM}	2.3	2.5	-	V _{RMS}	V _{IN} = 1V _{RMS} THD+N = 1% BW = 400-30kHz
Level	☆Output Gain	G _{H(OUT)}	0.5	2	3.5	dB	$G_{H(OUT)} = 20log(V_{OUT}/V_{IN})$

[☆]This Item is designated by ROHM only to discriminate between other items and it.

Unless otherwise specified, Ta=25°C, V_{CCL} =9V, V_{CCH} =11.5V, f=1kHz, V_{IN} =0.9 V_{RMS} , R_L =10k Ω , Input selector A, Input Gain 0dB, Gain Adjust +6dB, High-Voltage ON(High-Voltage mode1), LPF ON, Fader 0dB, Input point=INF1/INF2/INR1/INR2/INC/INS, Monitor point=OUTF1/OUTF2/OUTR1/OUTR2/OUTC/OUTS

~			Limit				
Block	Parameter	Symbol	Min	Тур	Max	Unit	Conditions
	Output Impedance	R _{OUT}	-	-	50	Ω	
Shift	☆Maximum Output Voltage	V _{OM}	3.2	3.4	-	V _{RMS}	V _{IN} =1V _{RMS} THD+N=1% BW=400-30kHz
Level	☆Output Gain	G _{H(OUT)}	2.6	4.6	6.6	dB	G _{H(OUT)} =20log(V _{OUT} /V _{IN})

[☆]This Item is designated by ROHM only to discriminate between other items and it.

Unless otherwise specified, Ta=25°C, V_{CCL} =9V, V_{CCH} =17V, f=1kHz, V_{IN} =0.9 V_{RMS} , R_L =10k Ω , Input selector A, Input Gain 0dB, Gain Adjust +6dB, High-Voltage ON(High-Voltage mode2), LPF ON, Fader 0dB, Input point=INF1/INF2/INR1/INR2/INC/INS, Monitor point=OUTF1/OUTF2/OUTR1/OUTR2/OUTC/OUTS

*			Limit				
Block	Parameter	Symbol	Min	Тур	Max	Unit	Conditions
	Output Impedance	R _{OUT}	-	-	50	Ω	
Shift	☆Maximum Output Voltage	V _{OM}	5.0	5.2	-	V _{RMS}	V _{IN} =1V _{RMS} THD+N=1% BW=400-30kHz
Level	☆Output Gain	G _{H(OUT)}	6.3	8.3	10.3	dB	G _{H(OUT)} =20log(V _{OUT} /V _{IN})

[☆]This Item is designated by ROHM only to discriminate between other items and it.

Unless otherwise specified, Ta=25°C, V_{CCL} =9V, V_{CCH} =17V, f=1kHz, V_{IN} =0.9 V_{RMS} , R_L =10k Ω , Input selector A, Input Gain 0dB, Gain Adjust +6dB, High-Voltage ON(High-Voltage mode2), LPF ON, Fader 0dB, Input point=INF1/INF2/INR1/INR2/INC/INS, Monitor point=OUTF1/OUTF2/OUTR1/OUTR2/OUTC/OUTS

				Limit			
Block	Parameter	Symbol	Min	Тур	Max	Unit	Conditions
	Maximum Boost Gain	G _{F BST}	21	23	25	dB	Fader Boost Gain = +23dB V_{IN} =0.1 V_{RMS} G_F =20log(V_{OUT}/V_{IN})- $G_{H(OUT)}$ Gain Adjust=0dB
	Channel Balance	СВ	-1.5	0	1.5	dB	CB = GV1-GV2
	Total Harmonic Distortion	THD+N	_	0.003	0.05	%	BW=400-30kHz Gain Adjust = 0dB
	Output Noise Voltage (Note1)	V_{NO1}	_	23	40	μV_{RMS}	$R_G = 0\Omega$ BW = IHF-A
	Residual Output Noise Voltage	V _{NOR}	_	10.5	20	μV _{RMS}	Fader Attenuation = $-\infty dB$ $R_G = 0\Omega$ BW = IHF-A
	Maximum Input Voltage	V _{IM}	2.0	2.1	_	V _{RMS}	V _{IM} at THD+N(V _{OUT})=1% BW=400-30kHz Gain Adjust = 0dB
Fader	Crosstalk Between Channels (Note1)	СТС	_	-100	-90	dB	$R_G = 0\Omega$ CTC=20log($V_{OUT}/V_{OUT'}$) BW = IHF-A
	Maximum Attenuation (Note1)	G _{F MIN}	_	-100	-90	dB	Fader Attenuation = -∞dB G _F =20log(V _{OUT} /V _{IN}) BW = IHF-A
	Gain Set Error	G _{F ERR}	-2	0	2	dB	Fader Boost Gain = +1 to +23dB
	Attenuation Set Error 1	G _{F ERR1}	-2	0	2	dB	Fader Attenuation = 0 to -15dB
	Attenuation Set Error 2	G _{F ERR2}	-3	0	3	dB	Fader Attenuation = -16 to -47dB
	Attenuation Set Error 3	G _{F ERR3}	-4	0	4	dB	Fader Attenuation = -48 to -79dB
	Power Supply Rejection Ratio	RR _{VCCL}	_	-70	-40	dB	V_{RR} =0.1 V_{RMS} f_{RR} =1 KHz RR_{VCCL} =20 $log(V_{OUT}/V_{CCL})$
	Tower supply regestion realio	RR _{VCCH}	_	-70	-40	dB	V_{RR} =0.1 V_{RMS} f_{RR} =1kHz RR_{VCCH} =20log(V_{OUT} / V_{CCH})
	Input Impedance	R _{IN_M}	70	100	130	kΩ	
bu	Maximum Input Voltage	V _{IM_M}	2.0	2.2	-	V _{RMS}	V _{IM_M} at THD+N(V _{OUT})=1% BW=400-30kHz Input point=MIN Front Mixing=OFF
Mixing	Maximum Attenuation (Note1)	G _{MX MIN}	-	-100	-85	dB	G _{MX} =20log(V _{OUT} /V _{IN}) BW=IHF-A Input point=MIN
	Mixing Gain	G _{MX}	-2	0	2	dB	Front Mixing=ON G _{MX} =20log(V _{OUT} /V _{IN})- G _{H(OUT)}
	Input Impedance	R _{IN_M}	70	100	130	kΩ	
Gain Adjust	Boost Gain	G _{F BST}	4	6	8	dB	Gain Adjust=6dB V _{IN} =0.1V _{RMS} G _F =20log(V _{OUT} /V _{IN})- G _{H(OUT)}
g	Channel Balance	СВ	-1.5	0	1.5	dB	CB = GV1-GV2

(Note1) VP-9690A (Average value detection, effective value display) filter by Panasonic is used for measurement. Input and output are in-phase.

Typical Performance Curve(s)

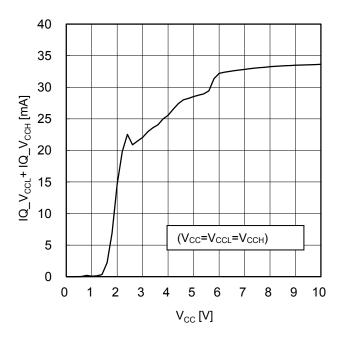


Figure 4. V_{CC} vs. $I_{Q_VCCL}+I_{Q_VCCH}$

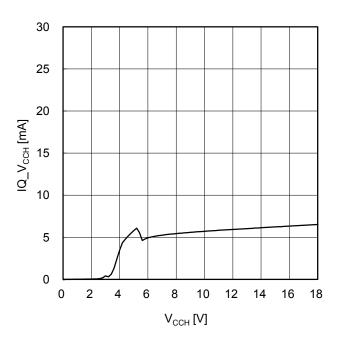


Figure 5. V_{CCH} vs. I_{Q_VCCH} (High-Voltage mode)

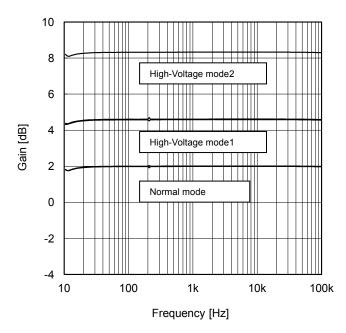


Figure 6. Gain vs. frequency (Normal / High-Voltage mode)

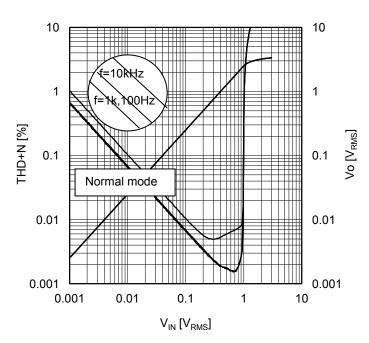
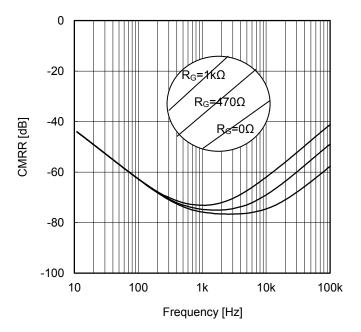


Figure 7. THD+N vs. Vin / Vo (Gain Adjust=+6dB)



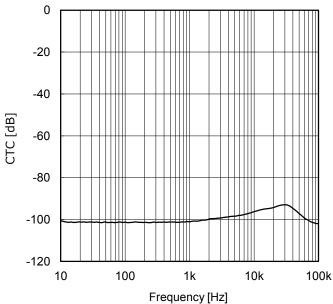


Figure 8. CMRR vs. frequency

Figure 9. Crosstalk (between Channels) vs. frequency

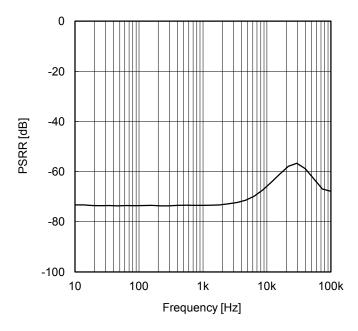


Figure 10. PSRR vs. frequency

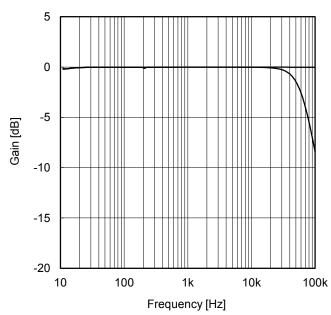


Figure 11. Gain(LPF ON/pass) vs. frequency

I²C-bus CONTROL SIGNAL SPECIFICATION

(1) Electrical specifications and timing for bus lines and I/O stages

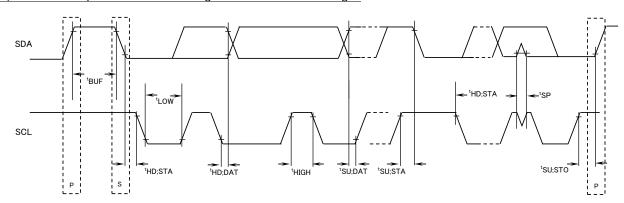


Figure 12. Definition of Timing on the I²C-bus

Table 1 Characteristics of the SDA and SCL bus lines for I²C-bus devices

	Parameter	Symbol	Fast-mode l	C-bus	Unit
	Parameter	Symbol	Min	Max	Unit
1	SCL Clock Frequency	f _{SCL}	0	400	kHz
2	Bus Free Time between STOP and START Condition	t _{BUF}	1.3	_	µsec
3	Hold Time (repeated) START condition. After this period, the first clock pulse is generated	t _{HD;STA}	0.6	_	µsec
4	LOW Period of the SCL Clock	t _{LOW}	1.3	_	µsec
5	HIGH Period of the SCL Clock	t _{HIGH}	0.6	_	µsec
6	Set-up Time for a Repeated START Condition	t _{SU;STA}	0.6	_	µsec
7	Data Hold Time	t _{HD;DAT}	0*	_	µsec
8	Data Set-up Time	t _{SU; DAT}	100	_	nsec
9	Set-up Time for STOP Condition	t _{su;sto}	0.6	_	µsec

All values referred to V_{IH} min. and V_{IL} max. Levels (see Table 2).

Table 2 Characteristics of the SDA and SCL I/O stages for I²C-bus devices

	Parameter	Symbol	Fast-mode l ²	C-bus	Unit
	Farameter	Symbol	Min	Max	o iii
10	LOW level input voltage: Fixed input levels	V _{IL}	-0.5	1	V
11	HIGH level input voltage: Fixed input levels	V _{IH}	2.3	-	V
12	Pulse width of spikes, which must be suppressed by the input filter.	t _{SP}	0	50	nsec
13	LOW level output voltage (open drain or open collector): At 3mA sink current	V _{OL1}	0	0.4	٧
14	Input current each I/O pin with an input voltage between 0.4V and 0.9 VDD max.	I _I	-10	10	μΑ

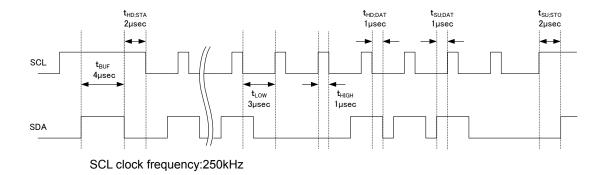


Figure 13. I²C-bus data transmission timing

(2) I²C-bus FORMAT

	MSB	LSB		MSB	LSB		MSB	LSE	3		
S	Slave	Address	Α	Select A	ddress	Α		Data	Α	Р	
1bit	8	bit	1bit	8	Bbit	1bit		8bit	1bit	1bit	
	S		= Sta	rt condition	(Recogniti	on of	start bit)				
	Slave	e Address	= Re	cognition of	slave addr	ess. 7	bits in up	per order are	e option	al.	
	The least significant bit is "L" due to write format.										
	Α		= AC	KNOWLED	GE bit (Re	cognit	ion of ack	nowledgeme	nt)		
	Selec	ct Address	= Sel	ection of re	gister that	contai	n data on	volume, bass	s and tr	eble se	ettings.
	Data		= Da	ta on every	volume an	d tone	to be sto	red in selecte	ed regis	ter.	
	Р		= Sto	p condition	(Recogniti	on of	stop bit)				

(3) I²C-bus Interface Protocol

1)Basic form

-	/									
	S	Slave Address	Α	Select	Address	Α	Dat	ta	Α	Р
_		MSB LS	3	MSB	LSB	М	SB	LSE	3	

2) Automatic increment (Select Address increases (+1) according to the number of data.)

S	Slave Ad	dress	Α	Select	Address	Α	Data1	Α	Data2	Α	 DataN	Α	Р
	MSB	LSB		MSB	LSB		MSB	LSB	MSB	LSB	MSB	L	SB

(Example) ① Data1 shall be set as data of address specified by Select Address.

- ② Data2 shall be set as data of address specified by Select Address +1.
- ③ DataN shall be set as data of address specified by Select Address +N-1.

3) Configuration unavailable for transmission (In this case, only Select Address1 is set.)

s	Slave Add	ress /	A Select	Address1	Α	Data	Α	Select Ad	dress 2	Α	Da	ıta	Α	Р
	MSB	LSB	MSB	LSB	М	SB LSE	3	MSB	LSB	М	SB	LSE	}	

(Note)If any data is transmitted as Select Address 2 next to data, It is recognized as data, not as Select Address 2.

(4) Slave address

MSB							LSB
A6	A5	A4	A3	A2	A1	A0	R/W
1	0	0	0	0	0	0	0

(5) Select Address & Data

Items	Select Address	MSB				Data			LSB		
items	(hex)	D7	D6	D5	D4	D3	D2	D1	D0		
Initial Setup 1	01	Advanced Switch ON/OFF	0	Time o	ed Switch of Input Fader	0	High-Voltage Mode Select	0	0		
Initial Setup 2	02	0	0	Sub S	elector	0	0 0 Rear Selector				
Input Selector	05	0	0	0	0		Input Selector				
Input Gain	06	0	0			Inp	Input Gain				
Fader 1ch Front	28	Fader Boost Gain / Attenuation									
Fader 2ch Front	29			Fa	der Boost (Gain / Atten	uation				
Fader 1ch Rear	2A			Fa	der Boost (Gain / Atten	uation				
Fader 2ch Rear	2B			Fa	ader Boost (Gain / Atten	uation				
Fader Center	2C			Fa	ader Boost (Gain / Atten	uation				
Fader Subwoofer	2D			Fa	ader Boost (Gain / Atten	uation				
LPF Setup Mixing ON/OFF	30	Front Mixing ON/OFF	LPF fc	0 0 0 Sub Main O Gain Gain Adjust Adjus							
Test Mode	F0	0	0	0	0	0	0	0	0		
System Reset	FE	1	0	0	0	0	0	0	1		

Notes on data format

- 1. "Advanced switch" function is available for the hatched parts on the above table.
- 2. In case of transferring data continuously, Select Address flows by Automatic Increment function, as shown below.

$$01(\text{hex}) \rightarrow 02(\text{hex}) \rightarrow 05(\text{hex}) \rightarrow 06(\text{hex}) \rightarrow 28(\text{hex}) \rightarrow 29(\text{hex}) \rightarrow 2A(\text{hex}) \rightarrow 2B(\text{hex}) \rightarrow 2C(\text{hex}) \rightarrow 2D(\text{hex}) \rightarrow 30(\text{hex}) \rightarrow 2D(\text{hex}) \rightarrow 2D($$

- 3. Input selector that is not corresponded for "Advanced switch" function, cannot reduce the noise caused when changing the input selector. Therefore, it is recommended to turn on mute when changing these settings.
- 4. In case of setting to infinite "-∞" by using Fader when input selector setting is changed, please consider "Advanced switch" time.

Explanation of each Select Address

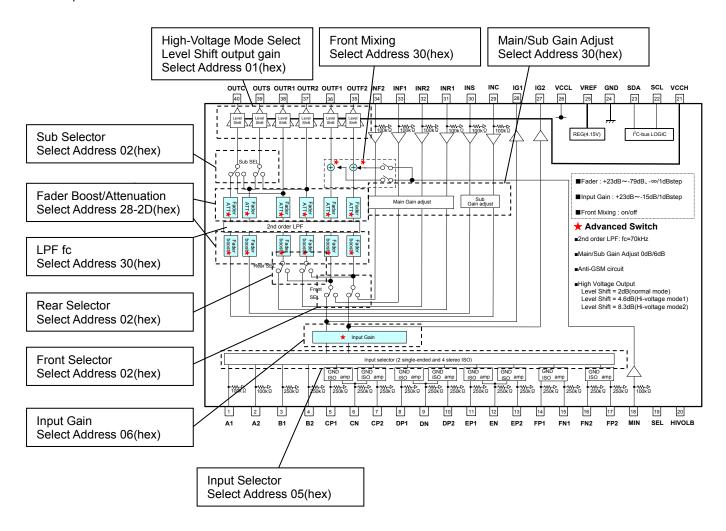


Figure 14. Block diagram for explanation of Select Address

Initial Condition,

1/0

Fixed value

Do not send the data not designated

Select Address 01 (hex)

٠.,	2017 taa1000 0 1 (110X)										
	Mode	MSB	High-Voltage Mode Select								
	Wode	D7	D6	D5	D4	D3	D2	D1	D0		
	High-Voltage mode2 (+8.3dB)					0	0	0	0		
	High-Voltage mode1 (+4.6dB)		0			0	1	U	U		

Mada	MSB	A	dvanced S	witch Time	of Input Gair	n/Fader ^{(Note}	e1,2)	LSB
Mode	D7	D6	D5	D4	D3	D2	D1	D0
4.7 msec			0	0				
7.1 msec		_	0	1	_			_
11.2 msec		0	1	0	0		0	U
14.4 msec			1	1				

(Note1) Advanced switch time is Typ value. Max value is 1.4 times of Typ value.
(Note2) If changing Advanced switch time while Advanced switch function is activated, Advance switch time is changed immediately.

Mode	MSB	MSB Advanced Switch ON/OFF ^(Note3)									
Iviode	D7	D6	D5	D4	D3	D2	D1	D0			
OFF	0	0			0		0	0			
ON	1	1			0		U	U			

(Note3) If Advanced switch ON/OFF is changed while Advanced switch function is activated, it will become effective from the next switching operation.

Select Address 02 (hex)

Mode	MSB	MSB Front Selector										
Iviode	D7	D6	D5	D4	D3	D2	D1	D0				
Front	0	0			0	0		0				
Inside Through	U	U			U	U		1				

Mode	MSB			Rea	r Selector			LSB
Wode	D7	D6	D5	D4	D3	D2	D1	D0
Rear	0	0			0	0	0	
Front Copy	U	U			U	U	1	

Mode ^(Note4)	MSB			Sub	Selector			LSB
Mode	D7	D6	D5	D4	D3	D2	D1	D0
OUTC(INS)/OUTS(INS)			0	0				
OUTC(INR1)/OUTS(INR2)	0	0	0	1	0	0		
OUTC (INC)/OUTS(INS)	U	0	1	0	U	U		
Prohibition			1	1				

(Note4) xxx(INxx): "xxx" means "Output terminal", "(INxx)" means "Output signal"

Initial Condition, 1/0 Fixed value Do not send the data not designated

Select Address 05(hex)

Mode	MSB			Input	Selector			LSB
iviode	D7	D6	D5	D4	D3	D2	D1	D0
Α					0	0	0	0
В					0	0	0	1
C single					0	0	1	0
D single					0	0	1	1
E single					0	1	0	0
F single					0	1	0	1
C diff.					0	1	1	0
D diff.	0	0	0	0	0	1	1	1
E diff.					1	0	0	0
F full-diff.					1	0	0	1
					1	0	1	0
Prohibition					1	0	1	1
Prombilion					:	:	:	:
					1	1	1	1

List of active input terminal when set input selector

Mode	Lch positive input terminal	Lch negative input terminal	Rch positive input terminal	Rch negative input terminal
Α	1pin(A1)	-	2pin(A2)	-
В	3pin(B1)	-	4pin(B2)	-
C single	5pin(CP1)	-	7pin(CP2)	-
D single	8pin(DP1)	-	10pin(DP2)	-
E single	11pin(EP1)	-	13pin(EP2)	-
F single	14pin(FP1)	-	17pin(FP2)	-
C diff.	5pin(CP1)	6pin(CN)	7pin(CP2)	6pin(CN)
D diff.	8pin(DP1)	9pin(DN)	10pin(DP2)	9pin(DN)
E diff.	11pin(EP1)	12pin(EN)	13pin(EP2)	12pin(EN)
F full-diff.	14pin(FP1)	15pin(FN1)	17pin(FP2)	16pin(FN2)

Initial Condition, 1/0 Fixed value Do not send the data not designated

Mode	MSB			Inp	out Gain			LSE
IVIOUE	D7	D6	D5	D4	D3	D2	D1	D0
			0	0	0	0	0	0
Prohibition			:	:	:	:	:	:
			0	0	1	0	0	0
+23dB			0	0	1	0	0	1
+22dB			0	0	1	0	1	0
+21dB			0	0	1	0	1	1
+20dB			0	0	1	1	0	0
+19dB			0	0	1	1	0	1
+18dB			0	0	1	1	1	0
+17dB			0	0	1	1	1	1
+16dB			0	1	0	0	0	0
+15dB			0	1	0	0	0	1
+14dB			0	1	0	0	1	0
+13dB			0	1	0	0	1	1
+12dB			0	1	0	1	0	0
+11dB			0	1	0	1	0	1
+10dB			0	1	0	1	1	0
+9dB			0	1	0	1	1	1
+8dB			0	1	1	0	0	0
+7dB			0	1	1	0	0	1
+6dB			0	1	1	0	1	0
+5dB			0	1	1	0	1	1
+4dB	0	0	0	1	1	1	0	0
+3dB			0	1	1	1	0	1
+2dB			0	1	1	1	1	0
+1dB			0	1	1	1	1	1
0dB			1	0	0	0	0	0
-1dB			1	0	0	0	0	1
-2dB			1	0	0	0	1	0
-3dB			1	0	0	0	1	1
-4dB			1	0	0	1	0	0
-5dB			1	0	0	1	0	1
-6dB			1	0	0	1	1	0
-7dB			1	0	0	1	1	1
-8dB			1	0	1	0	0	0
-9dB			1	0	1	0	0	1
-10dB			1	0	1	0	1	0
-11dB			1	0	1	0	1	1
-12dB			1	0	1	1	0	0
-13dB			1	0	1	1	0	1
-14dB			1	0	1	1	1	0
-15dB			1	0	1	1	1	1
			1	1	0	0	0	0
				:	:			:
Prohibition								

Initial Condition,

1/0 Fixed value

Do not send the data not designated

Select Address 28, 29, 2A, 2B, 2C, 2D (hex)

Boost & Attenuation	MSB		-	Fader Boo	st / Attenuat	ion		LSB
Boost & Attenuation	D7	D6	D5	D4	D3	D2	D1	D0
	0	0	0	0	0	0	0	0
Prohibition	0	0	0	0	0	0	0	1
Proffibilion	:	:	:	:	:	:	:	:
	0	1	1	0	1	0	0	0
+23dB	0	1	1	0	1	0	0	1
+22dB	0	1	1	0	1	0	1	0
+21dB	0	1	1	0	1	0	1	1
•			•					•
			-		•	•	•	
+10dB	0	1	1	1	0	1	1	0
+9dB	0	1	1	1	0	1	1	1
+8dB	0	1	1	1	1	0	0	0
+7dB	0	1	1	1	1	0	0	1
+6dB	0	1	1	1	1	0	1	0
+5dB	0	1	1	1	1	0	1	1
+4dB	0	1	1	1	1	1	0	0
+3dB	0	1	1	1	1	1	0	1
+2dB	0	1	1	1	1	1	1	0
+1dB	0	1	1	1	1	1	1	1
0dB	1	0	0	0	0	0	0	0
-1dB	1	0	0	0	0	0	0	1
-2dB	1	0	0	0	0	0	1	0
-3dB	1	0	0	0	0	0	1	1
: :	::	::		: :			::	::
-78dB	1	1	0	0	1	1	1	0
-79dB	1	1	0	0	1	1	1	1
	1	1	0	1	0	0	0	0
Prohibition	:	:	:	:	:	:	:	:
	1	1	1	1	1	1	1	0
-∞dB	1	1	1	1	1	1	1	1

Details of Fader Boost / Attenuation



(-ID)	D.7	DC	D.F.	D4	D0	D0	D4	D0	(-ID)	D.7	DC	Dr	D4	D0	D0	D4	D0
(dB)	D7	D6	D5	D4	D3	D2	D1	D0	(dB)	D7	D6	D5	D4	D3	D2	D1	D0
+23	0	1	1	0	1	0	0	1	-29	1	0	0	1	1	1	0	1
+22	0	1	1	0	1	0	1	0	-30	1	0	0	1	1	1	1	0
+21	0	1	1	0	1	0	1	1	-31	1	0	0	1	1	1	1	1
+20	0	1	1		1	1	0	0	-32		0	1	0	0	0	0	0
+19	0	1	1	0	1	1	0	1	-33	1	0	1	0	0	0	0	1
+18	0	1	1	0	1	1	1	0	-34	1	0	1	0	0	0	1	0
+17	0	1	1	0	0	0	0		-35 -36	1	0	1	0	0	0	0	1
+16 +15	0	1	1	1	0		0	0	-36	1	0	1	0	0	1	0	1
+15				1	_	0	1	0	-38	1	0	1		_	-	1	
	0	1	1		0	0							0	0	1		0
+13 +12	0	1	1	1	0	0	0	1	-39	1	0	1	0	0	1	0	1
+12	0	1	1	1	0	1	0	0	-40 -41	1	0	1	0	1	0	0	0
+10	0	1	1	1	0	1	1	0	-41 -42	1	0	1	0	1	0	1	0
+9	0	1	1	1	0	1	1	1	-42	1	0	1	0	1	0	1	1
+8	0	1	1	1	1	0	0	0	- 43 -44	1	0	1	0	1	1	0	0
+7	0	1	1	1	1	0	0	1	- 44 -45	1	0	1	0	1	1	0	1
+6	0	1	1	1	1	0	1	0	- 4 5	1	0	1	0	1	1	1	0
+5	0	1	1	1	1	0	1	1	-47	1	0	1	0	1	1	1	1
+4	0	1	1	1	1	1	0	0	-48	1	0	1	1	0	0	0	0
+3	0	1	1	1	1	1	0	1	-49	1	0	1	1	0	0	0	1
+2	0	1	1	1	1	1	1	0	-50	1	0	1	1	0	0	1	0
+1	0	1	1	1	1	1	1	1	-51	1	0	1	1	0	0	1	1
0	1	0	0	0	0	0	0	0	-52	1	0	1	1	0	1	0	0
-1	1	0	0	0	0	0	0	1	-53	1	0	1	1	0	1	0	1
-2	1	0	0	0	0	0	1	0	-54	1	0	1	1	0	1	1	0
-3	1	0	0	0	0	0	1	1	-55	1	0	1	1	0	1	1	1
-4	1	0	0	0	0	1	0	0	-56	1	0	1	1	1	0	0	0
-5	1	0	0	0	0	1	0	1	-57	1	0	1	1	1	0	0	1
-6	1	0	0	0	0	1	1	0	-58	1	0	1	1	1	0	1	0
-7	1	0	0	0	0	1	1	1	-59	1	0	1	1	1	0	1	1
-8	1	0	0	0	1	0	0	0	-60	1	0	1	1	1	1	0	0
-9	1	0	0	0	1	0	0	1	-61	1	0	1	1	1	1	0	1
-10	1	0	0	0	1	0	1	0	-62	1	0	1	1	1	1	1	0
-11	1	0	0	0	1	0	1	1	-63	1	0	1	1	1	1	1	1
-12	1	0	0	0	1	1	0	0	-64	1	1	0	0	0	0	0	0
-13	1	0	0	0	1	1	0	1	-65	1	1	0	0	0	0	0	1
-14	1	0	0	0	1	1	1	0	-66 67	1	1	0	0	0	0	1	0
-15 16	1	0	0	0	1	1	1	1	-67 69	1	1	0	0	0	0	1	1
-16	1	0	0	1	0	0	0	0	-68	1	1	0	0	0	1	0	0
-17 -18	1	0	0	1	0	0	0 1	0	-69 -70	1	1	0	0	0	1	0	0
-19	1	0	0	1	0	0	1	1	-70 -71	1	1	0	0	0	1	1	1
-19	1	0	0	1	0	1	0	0	-7 I -72	1	1	0	0	1	0	0	0
-21	1	0	0	1	0	1	0	1	-72 -73	1	1	0	0	1	0	0	1
-22	1	0	0	1	0	1	1	0	-73 -74	1	1	0	0	1	0	1	0
-23	1	0	0	1	0	1	1	1	-7 -7	1	1	0	0	1	0	1	1
-24	1	0	0	1	1	0	0	0	-76	1	1	0	0	1	1	0	0
-25	1	0	0	1	1	0	0	1	-77	1	1	0	0	1	1	0	1
-26	1	0	0	1	1	0	1	0	-78	1	1	0	0	1	1	1	0
-27	1	0	0	1	1	0	1	1	-79	1	1	0	0	1	1	1	1
-28	1	0	0	1	1	1	0	0	-∞	1	1	1	1	1	1	1	1

Initia

Initial Condition,

1/0 Fixed value

Do not send the data not designated

Select Address 30(hex)

Mode	MSB	MSB Main Gain Adjust									
Mode	D7	D6	D5	D4	D3	D2	D1	D0			
0dB			0	0	0	0		0			
+6dB			0	0	U	0		1			

Mode	MSB			Sub G	ain Adjust			LSB
Wode	D7	D6	D5	D4	D3	D2	D1	D0
0dB			0	0	0	0	0	
+6dB			0	U	U	U	1	

Mada	MSB												
Mode	D7	D6	D5	D4	D3	D2	D1	D0					
70kHz		0	0	0	0	0							
PASS		1	J	0	J	J							

Mode MSB			Front Mixing ON/OFF					LSB
Ivioue	D7	D6	D5	D4	D3	D2	D1	D0
OFF	0		0	0	0	0		
ON	1		U	U	U	U		

(6) About power on reset

It is possible for the reset circuit inside the IC to initialize when supply voltage is turned on. Please send data to all address as initial data when the supply is turned on, and turn on mute until all initial data are sent.

lt a ma	Cumahal		Limit		l lmi4	Condition	
Item	Symbol	Min	Тур	Max	Unit	Condition	
Rise time of VCCL	t _{RISE}	250	-	_	µsec	V _{CCL} rise time to 5V	
VCCL voltage of release power on reset	V _{POR}	_	4.1	_	V		

(7) About start-up and power off sequence on IC

By setting the terminal voltage of HIVOLB and SEL, it is possible to change the output gain. At the same time, output DC voltage will also be changed at each mode.

HIVOLB Terminal Voltage	High-Voltage
GND to 1.0V	ON
2.3V to VCCL	OFF

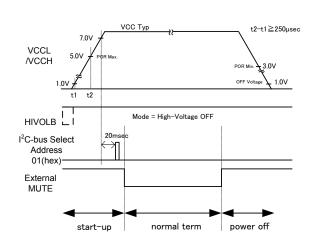
SEL Terminal Voltage	High-Voltage mode
GND to 0.5V	High-Voltage mode1
1.5V to VCCL	High-Voltage mode2

Please set HIVOLB terminal voltage between the ranges showed by the above tables. If HIVOLB terminal is open, the terminal voltage will be set to 5V due to the pull-up voltage inside the IC. In this case, the IC will be set to "High-Voltage OFF" mode. SEL terminal is 4.15V due to the pull-up voltage inside the IC.

Output DC voltage and Output gain, that are changed by the combination of "HIVOLB" terminal and "SEL" terminal shows as the following table.

VCCH Supplied Voltage	9 \	/	11.5 V	17 V
HIVOLB Terminal Voltage	5 \ (High-Volta		0\ (High-Volt	
SEL Terminal Voltage	0V Open (4.15 V) (High-Voltage mode1) (High-Voltage mode2		0 V (High-Voltage mode1)	Open (4.15 V) (High-Voltage mode2)
Output DC Bias Voltage	4.35 V 2 dB		5.6 V	8.35 V
Level Shift Output gain			4.6 dB	8.3 dB

If HIVOLB terminal voltage is changed during its operation, Output DC voltage will be also changed shown as above. For reducing these variations, turn the power on after setting the status of the HIVOLB and SEL terminal according to the output gain. The start-up and power off sequence is shown next.



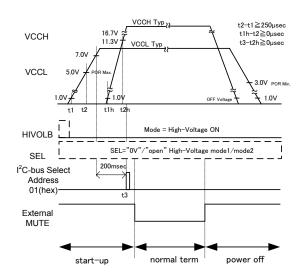


Figure 15. Normal mode(High-Voltage OFF) operation

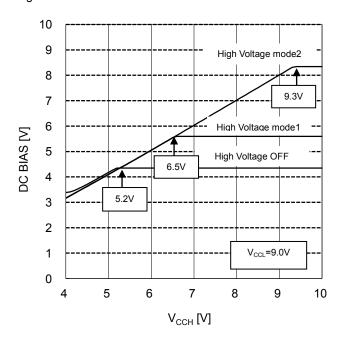
Figure 16. High-Voltage mode operation (High-Voltage mode1 and mode2 common)

HIVOLB in the figure above is used to select the Output gain. This IC will become active-state by sending data of Select Address 01(hex) on I²C-bus after 20msec from that VCCL reaches over 7.0V. High-Voltage Output gain is selected by setting SEL terminal voltage and sending I²C-bus data. Therefore, this command must always be sent in the start-up sequence. In addition, "External MUTE" in the figure above is the recommended period that the muting is activated from outside the IC. In addition, the starting sequence of VCCL and VCCH does not have the limit, but please start VCCL earlier to reduce a pop noise.

For HIVOLB terminal, there is countermeasure taken for protection from voltage spikes. But, please take care that the output DC voltage may fluctuate, if the period of voltage spike is over 50nsec.

(8) About relations of power supply voltage and the DC-bias voltage

Output DC-bias voltage is decided by the regulator that is embedded in this IC, DC-bias does not fluctuate up to a constant level even if power supply voltage is lowered. The following graphs show the relationship between DC-bias voltage and power supply voltage.





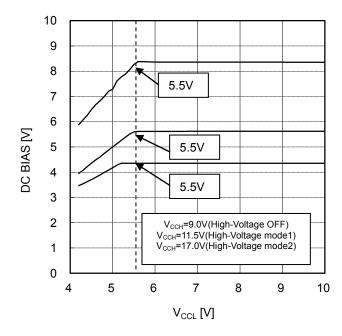


Figure 18. V_{CCL} vs DC Bias

About advanced switch circuit

- [1] Advanced switch technology
- 1-1. Advanced switch effects

Advanced switch technology is ROHM original technology that can prevent from switching pop noise. If changing the gain setting (for example Fader) immediatery, the audible signal will become discontinuously and pop noise will be occured. This Advanced switch technology will prevent this discontinuous signal by completing the signal waveform and will significantly reduce the noise.

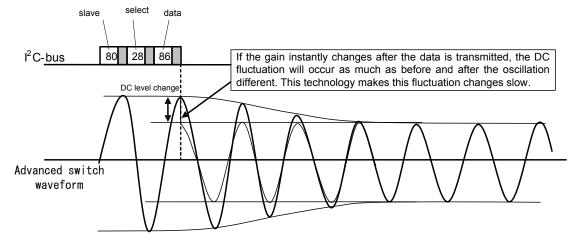


Figure 19. Advanced switch waveform

This Advanced switch circuit will start operating when the data is transmitted from microcontroller. Advanced switch waveform is shown as the figure above. For preventing switching noise, This IC will operate optimally by internal processing after the data is transmitted from microcontroller.

However, sometimes the switching waveform is not like the intended form depends on the transmission timing. Therefore, below is the example of the relationship between the transmission timing and actual switching time. Please consider this relationship for the setting.

1-2. The kind of the Transferring Data

- Data setting that is not corresponded to Advanced switch (<u>(5)Select Address & data</u> Data format without hatching)
 There is no particular rule about transferring data.
- Data setting that is corresponded to Advanced switch (Note1)
 ((5)Select Address & data Data format with hatching)
 There is no particular rule about transferring data, but Advanced switch must follow the switching sequence as mentioned in [2] as follows.

(Note1) The blocks that are corresponded Advanced switch are "Input Gain", "Fader" and "Front Mixing ON/OFF" (In detail, please refer to (5) Select Address & data).

[2] Data transmission that is corresponded to Advanced switch

2-1. Switching time of Advanced switch

Switching time includes [Twait(Wait time)], [Tsft(A \rightarrow B switching time)] and [Tsft(B \rightarrow A switching time)]. 25msec is needed per 1 switching. (Tsoft = Twait + 2 * Tsft, Twait=2.3msec, Tsft=11.2msec)

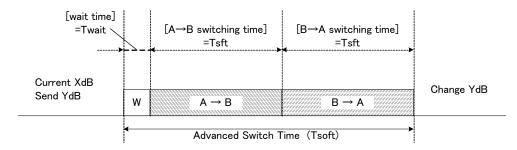


Figure 20. About Advanced switching time

In the figure above, Start/Stop state is expressed as "A" and temporary state is expressed as "B".

The switching sequence of Advanced switch consists of the cycle "A(start) \rightarrow B(temporary) \rightarrow A(stop)". Therefore, switching sequence will not stop at B state.

For example, switching is performed from A(Initial gain) \rightarrow B(set gain) \rightarrow A(set gain) when switching from initial gain to set gain. And switching time (Tsft) of A \rightarrow B or B \rightarrow A are equal.

2-2. Explanation on data transmission's timing and switching operation.

The following examples show the timing chart from data transmission to starting of switching. Definition of example expression :

F1=Fader 1ch Front, F2=Fader 2ch Front, R1=Fader 1ch Rear, R2=Fader 2ch Rear

C=Fader Center, S=Fader Subwoofer, MIX=Front Mixing

■ Transmission example 1

This is an example when transmitting data in same block with "enough interval for data transmission". (enough interval for data transmission : 1.4 x Tsoft * "1.4" includes tolerance margin.)

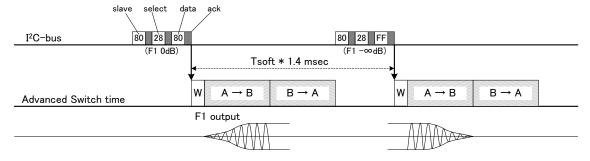


Figure 21. Transmission example 1

- Transmission example 2
- This is an example when the transmission interval is not enough (smaller than "Transmission example 1").
- When the data is transmitted during first switching operation, the second data will be reflected after the first switching operation. In this case, there is no wait time (Twait) before the second switching operation.

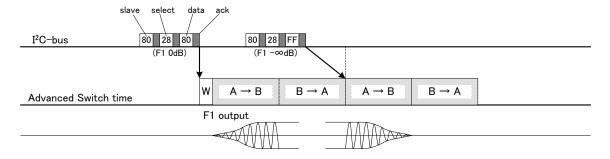


Figure 22. Transmission example 2

■ Transmission example 3

This is an example when transmission interval is smaller than "Transmission example 2"). When the data is transmitted during the first switching operation, and transmission timing is just during $A \rightarrow B$ switching operation, the second data will be reflected at $B \rightarrow A$ switching term in case of Fader.

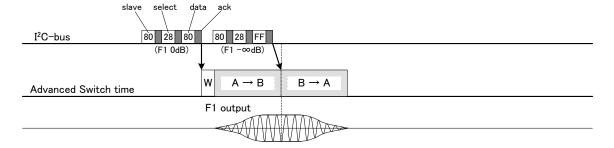


Figure 23. Transmission example 3

Please take care as follows when transmitting data to multiple channels.

It is possible that Lch and Rch in same block(Front/Rear/Center,Subwoofer) can be switched at the same timing. For example, if the data transmission is set as the figure below, F1 and F2 can be switched at the same timing. (Data ①is sent for F1 (Lch) and data ②is sent for F2 (Rch).)

Twait (designed to 2.3 msec) is the wait time for starting switching.

Twait may change from 1.2msec (Min.) to 4.6msec (Max.) by considering tolerance margin.

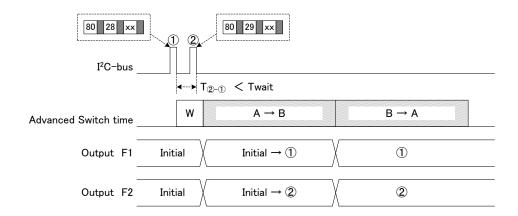


Figure 24. The operation during multi-channels (Lch, Rch) data transmission (smaller than Twait interval).

Next, if data ②is not transmitted during the Twait, the switching operation will be as the figure below.

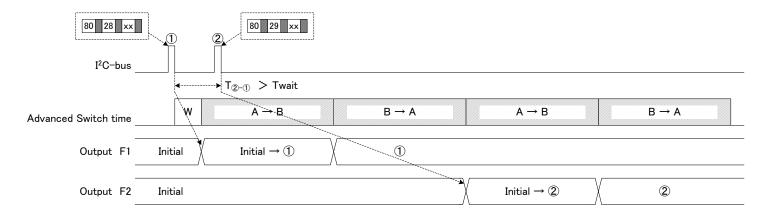


Figure 25. The operation during multiple channels (Lch, Rch) data transmission (larger than Twait interval).

2-3. Multiple blocks data transmission timing and switching operation.

In case the data is transmitted to multiple blocks, the processing is performed internally by BS (Block state) unit. Starting order of Advanced switch is determined by BS unit.

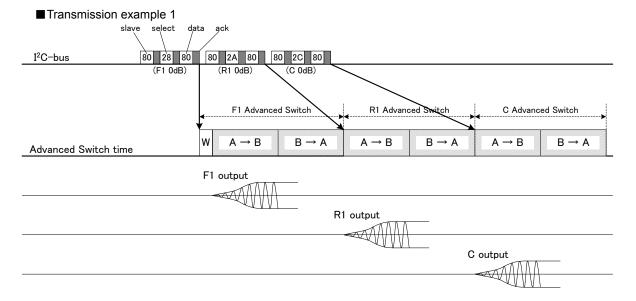


Figure 26. Multi-blocks data transmission timing

There are no timing regulations of I^2C -bus data transmission. But next switching will start after the end of the current switching. The timing of Advanced switch is depended, not on the order of data transmission, but on the order of the figure below. The blocks in the same group (For example, BS1, BS3) can start switching at the same time.

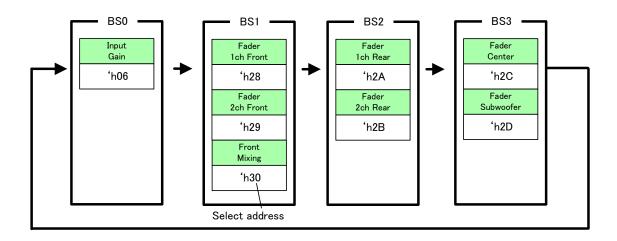


Figure 27. The turn of Advanced switch start

■ Transmission example 2

In case of that the transmission order is different with actual switching order.

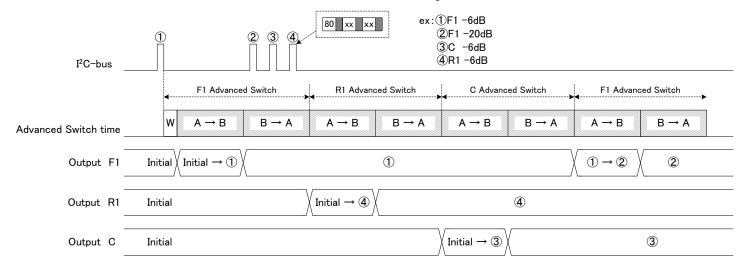


Figure 28. In case of the transmission order is different with actual switching order

If the data of Front/Rear/Center setting is transmitted during the switching of Front, Rear and Center switching have priority over Front switching. In order to proceeding the switching starts as the data transmission order, please transmit the next data after the end of current switching.

■Transmission example 3

If Refresh data that is same as current setting is transmitted, gain switching operation will not start.

The below figure shows the case of transmitting data after Refresh data.

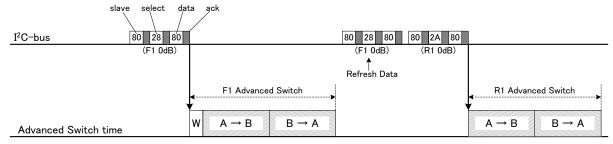


Figure 29. In case of the transmission of Refresh data

Application Example

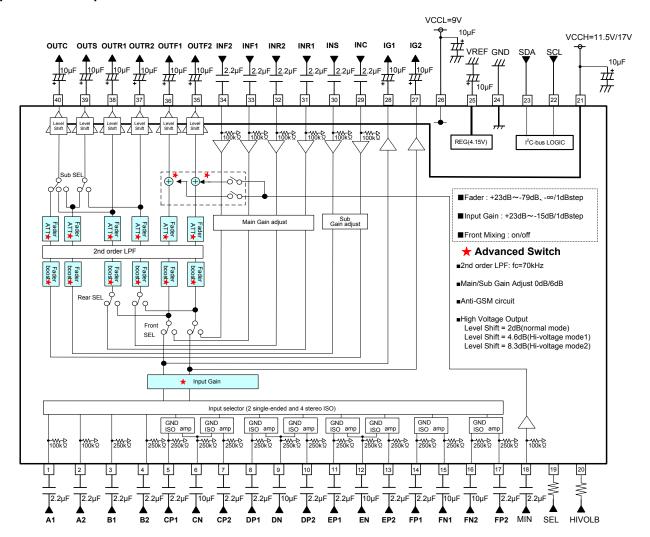


Figure 30. Application Example

UNIT RESISTANCE: Ω CAPACITANCE: F

Notes on wiring

- ①Please connect the decoupling capacitor of a power supply as close as possible to GND.
- 2 Lines of GND shall be one-point connected.
- ③Wiring pattern of Digital unit shall be away from that of analog unit and crosstalk shall not be acceptable.
- (4) Lines of SCL and SDA of I²C-bus shall not be parallel if possible. The lines shall be shielded, if they are adjacent to each other
- ⑤Lines of analog input shall not be parallel if possible. The lines shall be shielded, if they are adjacent to each other.

O Equivalenc		<u> </u>		
Terminal	Terminal	Terminal	Equivalence Circuit	Terminal Description
No	Name	Voltage	4	A terminal for signal input
1	A1	4.15V	VCCL	
2	A2		0	The input impedance is 100kΩ(Typ).
29	INC		≰ ∜⊜	
30	INS			
31	INR1			
32	INR2		100kΩ	
33	INF1		▼ 4.15V	
34	INF2		GND V 4.15V	
18	MIN			
3 4 5	B1 B2 CP1	4.15V	Voci	Input terminal Single/Differential mode is selectable. The input impedance is $250k\Omega(Typ)$.
6	CN		VCCL •	The input impedance is 200kg (Typ).
7	CP2			
8	DP1		l T l	
9	DN			
10	DP2		\$ 250kΩ	
11	EP1		↑ ` ' '	
12	EN		GND	
13	EP2		0	
14	FP1			
15	FN1			
16	FN2			
17	FP2			Input gain output terminal
27	IG2	4.15V	VCCL	mpat gam output terminar
28	IG1		GND O	
35	OUTF2	(1) 4.35V	VCCH	Fader Output terminal (1) Normal mode : 4.35V
36	OUTF1	(2) 5.6V		(2) High-Voltage mode1 : 5.6V
37	OUTR2	(3) 8.35V	│ ╁	(3) High-Voltage mode2 : 8.35V
38	OUTR1			
39	OUTS			
40	OUTC			
			GND	

The figures in the pin explanation and input/output Equivalence circuit are reference values, it doesn't guarantee exact values.

Terminal	Terminal	Terminal	Equivalence Circuit	Terminal Description
No	Name	Voltage	VCCL	Output gain control terminal
20	HIVOLB	5V	5V	Low(0V supply) : High-Voltage ON High(terminal open) : High-Voltage OFF
21	VCCH	17/11.5/9V		Power supply terminal.
26	VCCL	9V	Voc	T
22	SCL		VCCL I.65V	Terminal for clock input of I ² C-bus communication. Note: When this pin is shorted to next pin(VCCH), it may result in property degradation and destruction of the device.
23	SDA		VCCL O 1.65V	Terminal for data input of I ² C-bus communication.
24	GND	0V		Ground terminal.
25	VREF	4.15V	VCCL O	BIAS terminal.
			12.5kΩ 4.15V	Voltage for reference bias of analog signal system. The simple precharge circuit and simple discharge circuit for an external capacitor are built in.
19	SEL	4.15V	VCCL O	High Voltage Output Select terminal
			GND 250kΩ	Low(0V supply) : High-Voltage mode1 High(terminal open) : High-Voltage mode2

The figures in the pin explanation and input/output Equivalence circuit are reference values, it doesn't guarantee exact values.

Application Information

1) Absolute maximum rating voltage

When voltage is impressed to VCCL/VCCH exceeding absolute maximum rating voltage, circuit current increases rapidly and it may result in property degradation and destruction of a device. When impressed by a VCCL terminal (26pin) especially by surge examination etc., even if it includes an of operation voltage +surge pulse component, be careful not to impress voltage (about 14V) much higher than absolute maximum rating voltage. And, be careful that there is no more than 18V on the VCCH terminal (21pin).

2) About a signal input part

In the signal input terminal, the value of the input coupling capacitor C(F) should be sufficient—to match the value of input impedance $R_{IN}(\Omega)$ inside the IC. The first HPF characteristic of CR is as shown below.

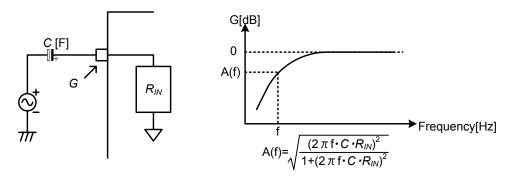


Figure 31. Input Equivalence Circuit

3) About output load characteristics

The usages of load for output are below (reference). Please use the load more than 10 k Ω (Typ).

Out	Output terminal									
	Terminal	Terminal	Terminal	Terminal	Terminal	Terminal	Terminal	Terminal		
	No.	Name	No.	Name	No.	Name	No.	Name		
	28	IG1	36	OUTF1	38	OUTR1	40	OUTC		

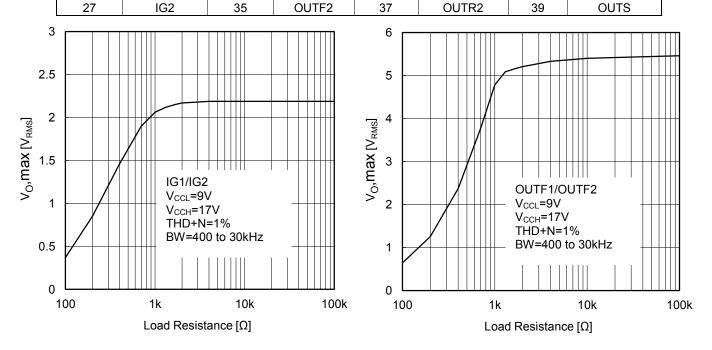


Figure 32. Output load characteristic at V_{CCL}=9V, V_{CCH}=17V(Reference)

4) About HIVOLB terminal(20pin) when power supply is off

Any voltage shall not be supplied to HIVOLB terminal (20pin) when power-supply is off. Please insert a resistor (about 2.2kΩ) to HIVOLB terminal in series, in case voltage is supplied to HIVOLB terminal. (Please refer Application Circuit Diagram.)

5) About signal input terminals

Because the inner impedance of the terminal becomes 100 k Ω or 250 k Ω when the signal input terminal makes a terminal open, the plunge noise from outside sometimes becomes a problem. When there is an unused signal input terminal, design so it is shorted to ground.

6) About changing gain of Input Gain and Fader Volume

When increasing the input gain and fader volume, especially those exceeding 20dB, the switching pop noise sometimes becomes big. In this case, we recommend changing the gain in 1 dB steps, without abruptly changing the gain at once. Also, the pop noise can sometimes be reduced by increasing the advanced switch time.

7) About inter-pin short to VCCH

VCCH terminal(21pin) is assumed at applied high voltage(Max 17.8V) for $5.2V_{RMS}$ (Max) output. And so, avoid short between VCCH and SCL. When Inter-pin shorts occur, circuit current increases rapidly, and it may result in property degradation and destruction of a device.

Operational Notes

Reverse Connection of Power Supply

Connecting the power supply in reverse polarity can damage the IC. Take precautions against reverse polarity when connecting the power supply, such as mounting an external diode between the power supply and the IC's power supply pins.

Power Supply Lines

Design the PCB layout pattern to provide low impedance supply lines. Separate the ground and supply lines of the digital and analog blocks to prevent noise in the ground and supply lines of the digital block from affecting the analog block. Furthermore, connect a capacitor to ground at all power supply pins. Consider the effect of temperature and aging on the capacitance value when using electrolytic capacitors.

3. Ground Voltage

Ensure that no pins are at a voltage below that of the ground pin at any time, even during transient condition.

4. Ground Wiring Pattern

When using both small-signal and large-current ground traces, the two ground traces should be routed separately but connected to a single ground at the reference point of the application board to avoid fluctuations in the small-signal ground caused by large currents. Also ensure that the ground traces of external components do not cause variations on the ground voltage. The ground lines must be as short and thick as possible to reduce line impedance.

5. Thermal Consideration

Should by any chance the maximum junction temperature rating be exceeded the rise in temperature of the chip may result in deterioration of the properties of the chip. In case of exceeding this absolute maximum rating, increase the board size and copper area to prevent exceeding the maximum junction temperature rating.

6. Recommended Operating Conditions

These conditions represent a range within which the expected characteristics of the IC can be approximately obtained. The electrical characteristics are guaranteed under the conditions of each parameter.

7. Inrush Current

When power is first supplied to the IC, it is possible that the internal logic may be unstable and inrush current may flow instantaneously due to the internal powering sequence and delays, especially if the IC has more than one power supply. Therefore, give special consideration to power coupling capacitance, power wiring, width of ground wiring, and routing of connections.

8. Operation Under Strong Electromagnetic Field

Operating the IC in the presence of a strong electromagnetic field may cause the IC to malfunction.

9. Testing on Application Boards

When testing the IC on an application board, connecting a capacitor directly to a low-impedance output pin may subject the IC to stress. Always discharge capacitors completely after each process or step. The IC's power supply should always be turned off completely before connecting or removing it from the test setup during the inspection process. To prevent damage from static discharge, ground the IC during assembly and use similar precautions during transport and storage.

Operational Notes - continued

10. Inter-pin Short and Mounting Errors

Ensure that the direction and position are correct when mounting the IC on the PCB. Incorrect mounting may result in damaging the IC. Avoid nearby pins being shorted to each other especially to ground, power supply and output pin. Inter-pin shorts could be due to many reasons such as metal particles, water droplets (in very humid environment) and unintentional solder bridge deposited in between pins during assembly to name a few.

11. Unused Input Pins

Input pins of an IC are often connected to the gate of a MOS transistor. The gate has extremely high impedance and extremely low capacitance. If left unconnected, the electric field from the outside can easily charge it. The small charge acquired in this way is enough to produce a significant effect on the conduction through the transistor and cause unexpected operation of the IC. So unless otherwise specified, unused input pins should be connected to the power supply or ground line.

12. Regarding the Input Pin of the IC

This monolithic IC contains P+ isolation and P substrate layers between adjacent elements in order to keep them isolated. P-N junctions are formed at the intersection of the P layers with the N layers of other elements, creating a parasitic diode or transistor. For example (refer to figure below):

When GND > Pin A and GND > Pin B, the P-N junction operates as a parasitic diode. When GND > Pin B, the P-N junction operates as a parasitic transistor.

Parasitic diodes inevitably occur in the structure of the IC. The operation of parasitic diodes can result in mutual interference among circuits, operational faults, or physical damage. Therefore, conditions that cause these diodes to operate, such as applying a voltage lower than the GND voltage to an input pin (and thus to the P substrate) should be avoided.

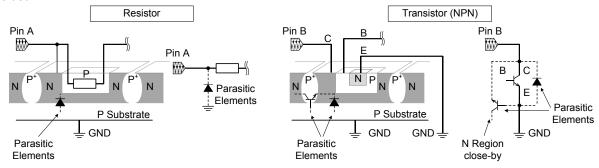
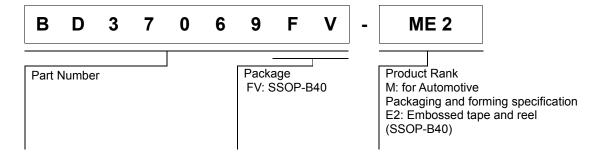
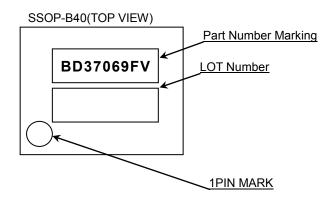


Figure 33. Example of monolithic IC structure

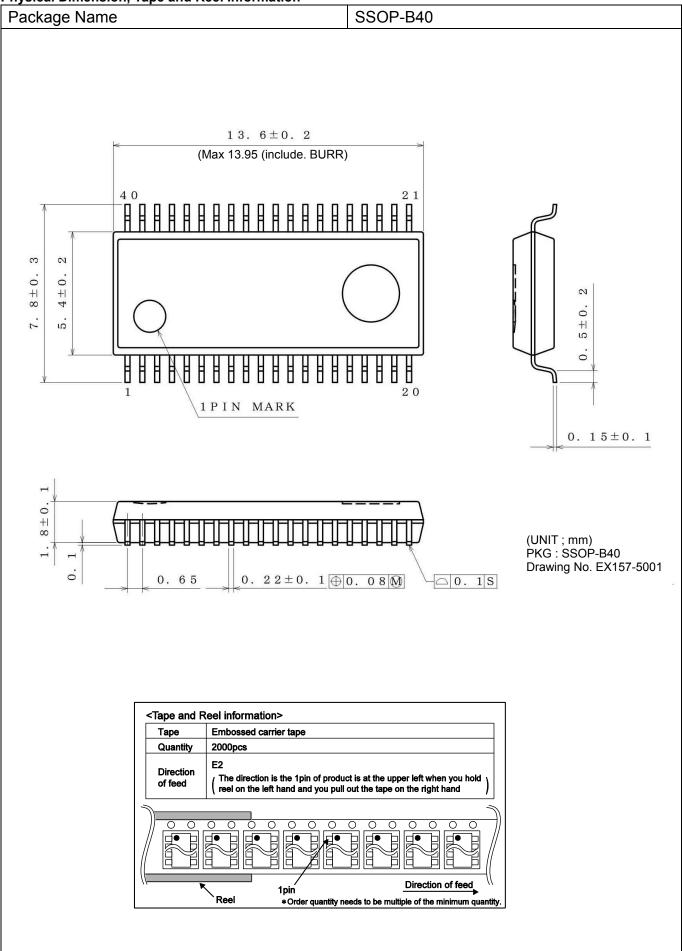
Ordering Information



Marking Diagram



Physical Dimension, Tape and Reel Information



Revision History

Date	Revision	Changes	
12.MAY.2016	001	New Release	

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CLASSIV	CLASSIII	CLASSⅢ	CLASSIII

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